

EE 330

Lecture 7

Propagation Delay

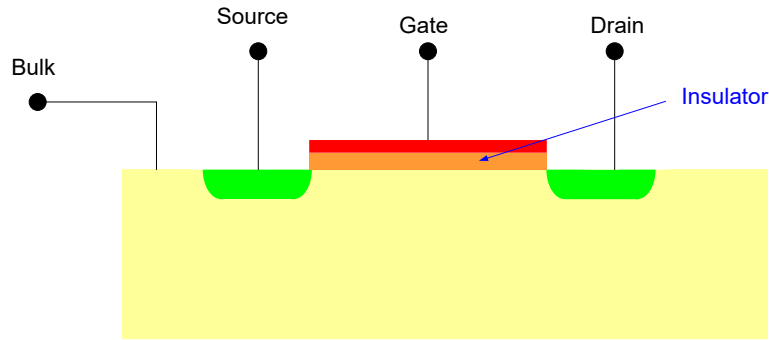
Stick Diagrams

Technology Files

Review from Last Time

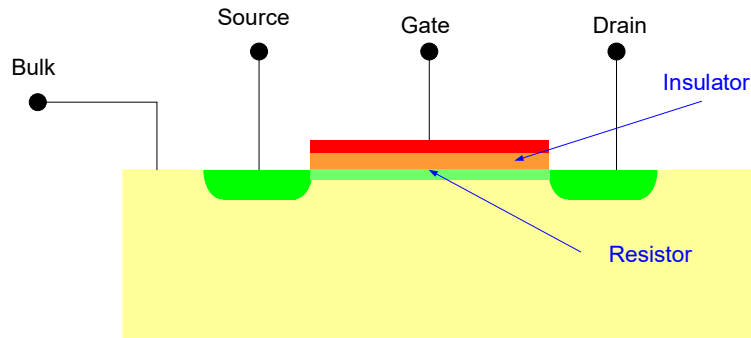
MOS Transistor

Qualitative Discussion of n-channel Operation



n-channel MOSFET

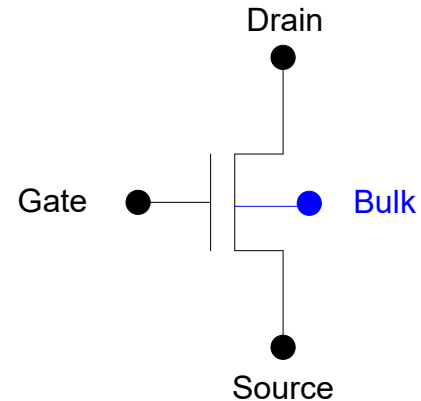
For V_{GS} small



n-channel MOSFET

For V_{GS} large

- Region under gate termed the “channel”
- When “resistor” is electrically created, region where it resides in channel is termed an “inversion region”

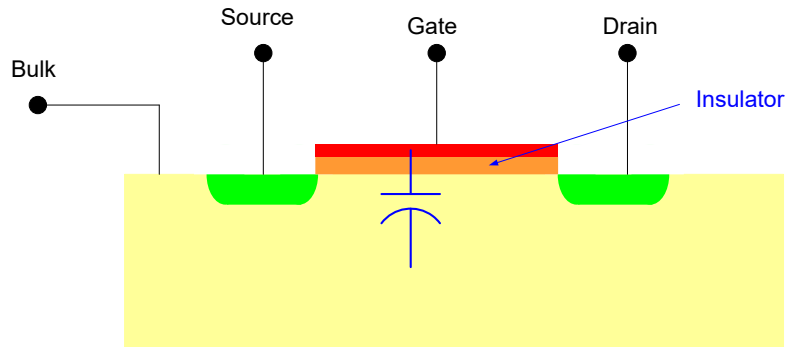


MOSFET actually 4-terminal device

Review from Last Time

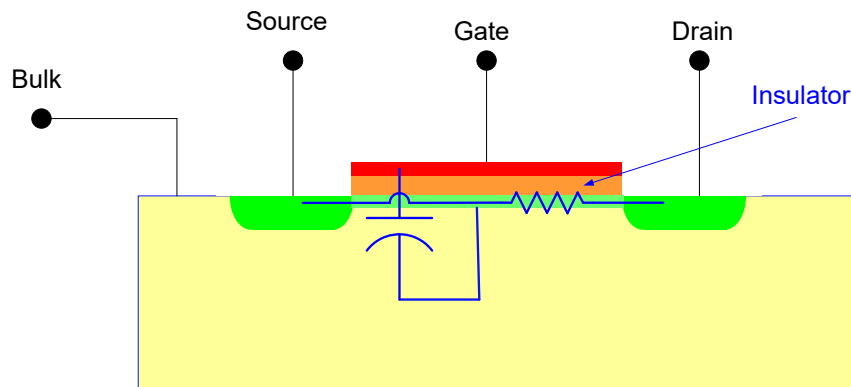
MOS Transistor

Qualitative Discussion of n-channel Operation



n-channel MOSFET

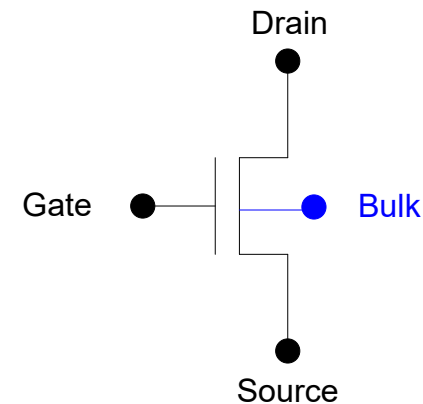
For V_{GS} small



n-channel MOSFET

For V_{GS} large

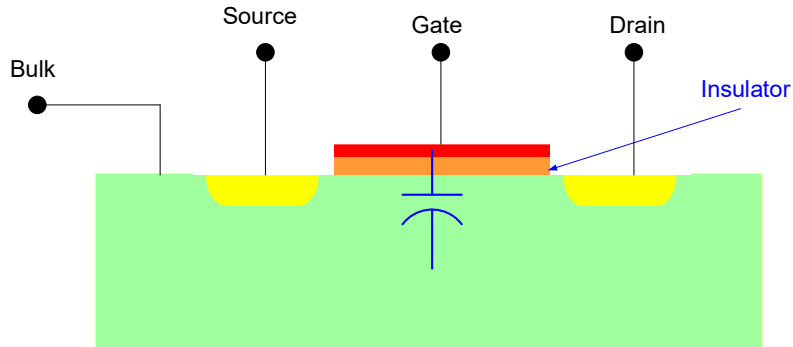
- Electrically created inversion layer forms a “thin “film” resistor
- Capacitance from gate to channel region is distributed
- Lumped capacitance much easier to work with



Review from Last Time

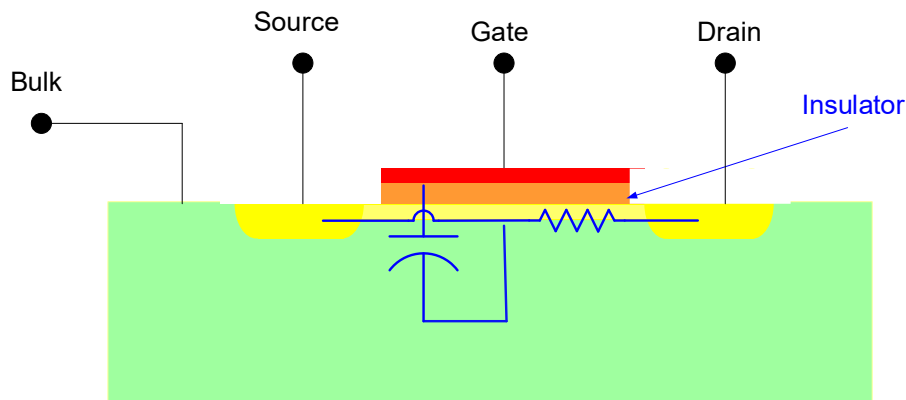
MOS Transistor

Qualitative Discussion of p-channel Operation

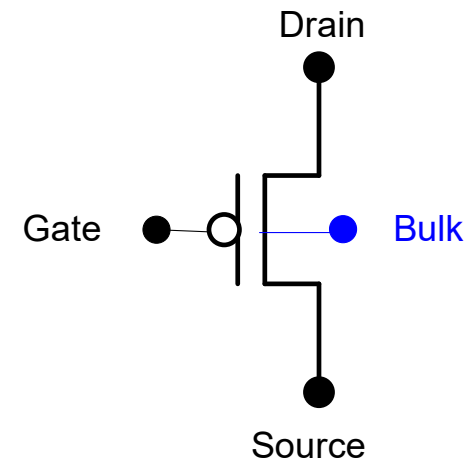


p-channel MOSFET

For $|V_{GS}|$ small



p-channel MOSFET



For $|V_{GS}|$ large

- Electrically created inversion layer forms a “thin “film” resistor
- Capacitance from gate to channel region is distributed
- Lumped capacitance much easier to work with

Improved Switch-Level Model



C_{GS} and R_{SW} dependent upon device sizes and process

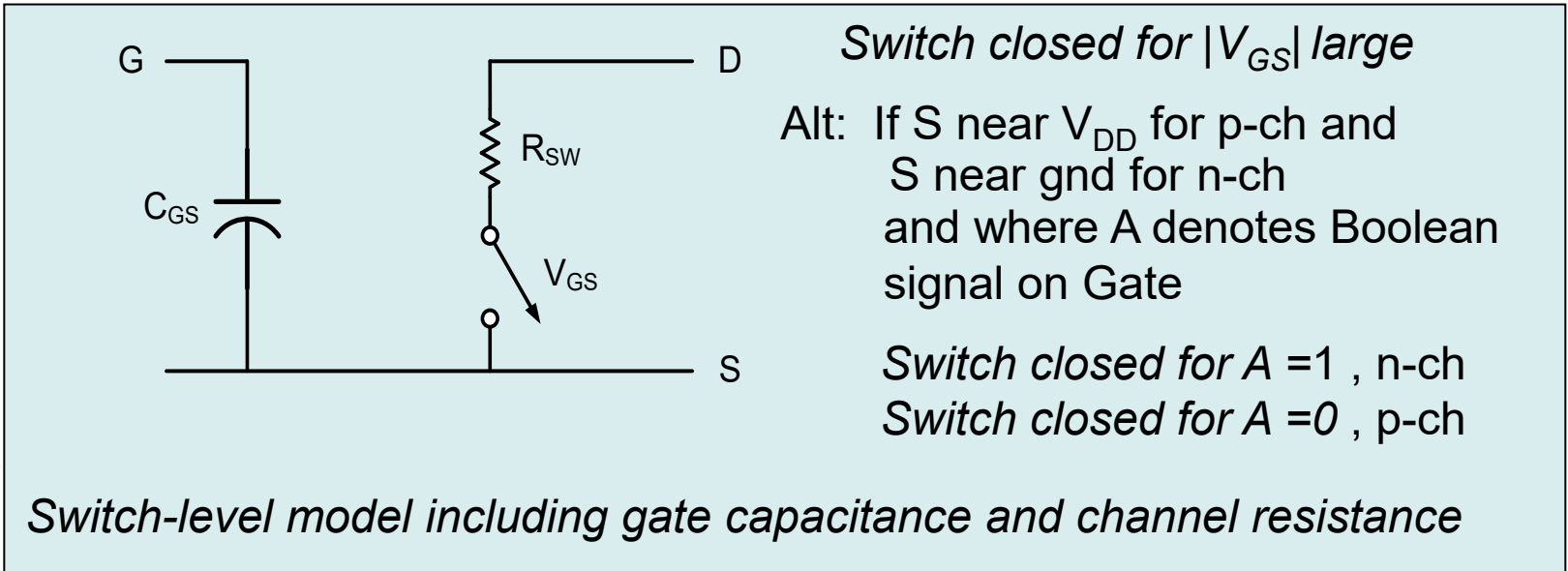
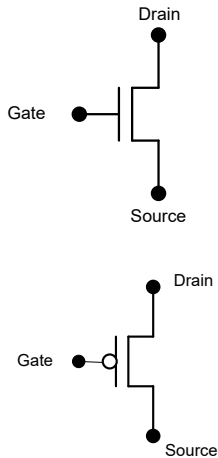
For minimum-sized devices in a 0.5 μ process with V_{DD} =5V

$$\mathbf{C}_{GS} \cong 1.5\text{fF} \qquad \mathbf{R}_{sw} \cong \left. \begin{array}{l} 2\text{K}\Omega \text{ n-channel} \\ 6\text{K}\Omega \text{ p-channel} \end{array} \right\}$$

Considerable emphasis will be placed upon device sizing to manage C_{GS} and R_{SW}

Review from Last Time

Improved Switch-Level Model



C_{GS} and R_{SW} dependent upon device sizes and process

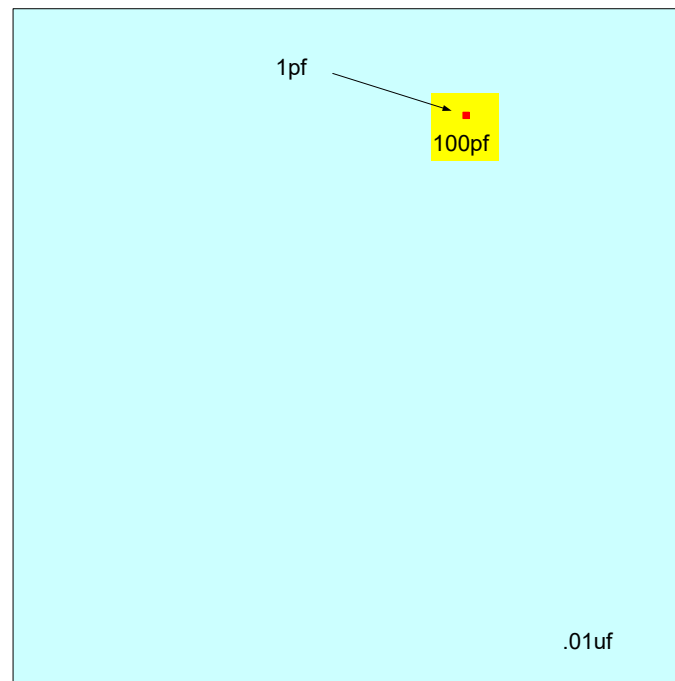
Will assume for minimum-sized devices in a 0.18u process with $V_{DD}=2V$

$$C_{GS} \cong 1.5\text{fF} \quad R_{sw} \cong \left. \begin{array}{l} 2\text{K}\Omega \text{ n-channel} \\ 6\text{K}\Omega \text{ p-channel} \end{array} \right\}$$

Considerable emphasis will be placed upon device sizing to manage C_{GS} and R_{SW}

Review from Last Time

Is a capacitor of 1.5fF small enough to be neglected?



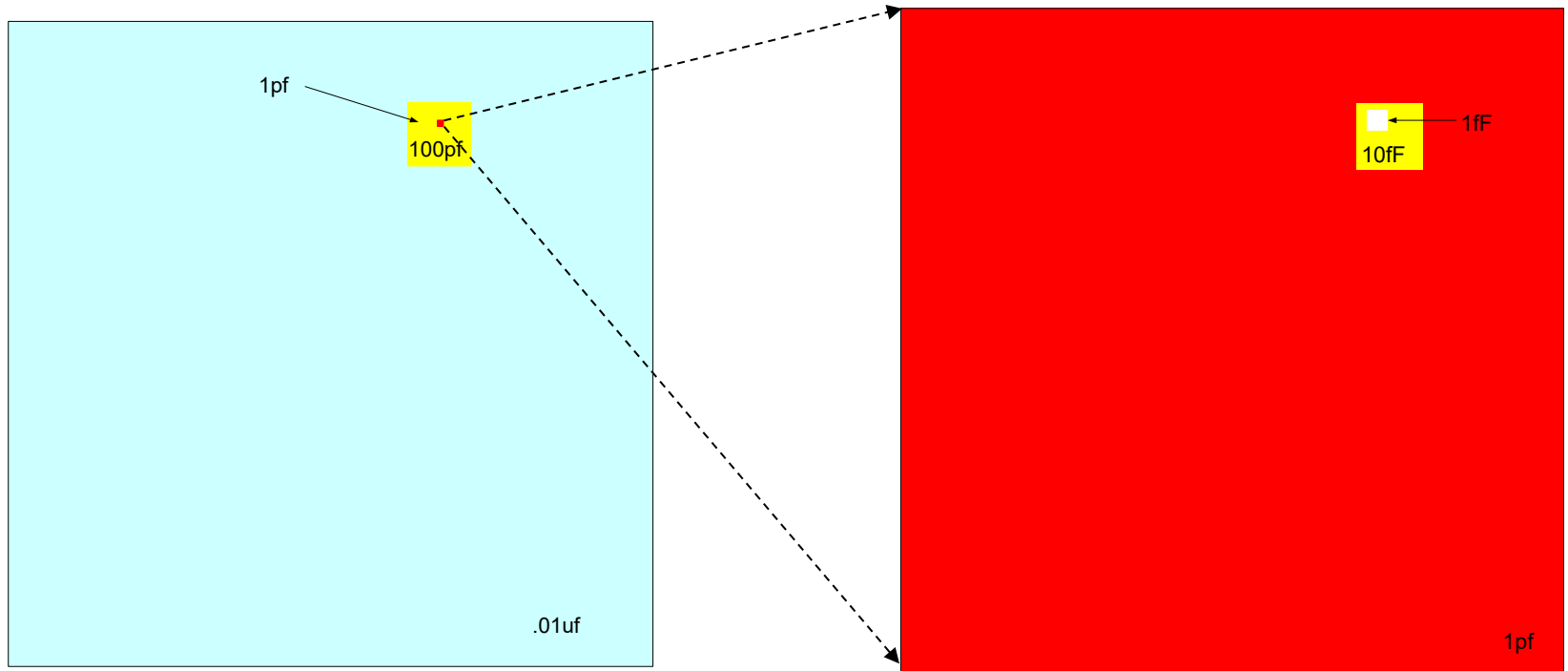
From EE 201 Parts Kit

Capacitors (Farads)		
100p	3	
470p	3	
0.001u	3	2
0.0047u	3	2
0.01u	3	
0.047u	3	
0.1u	3	1
0.47u	3	
1u	3	
10u	3	
100u	3	

Area allocations shown to relative scale:

Review from Last Time

Is a capacitor of 1.5fF small enough to be neglected?

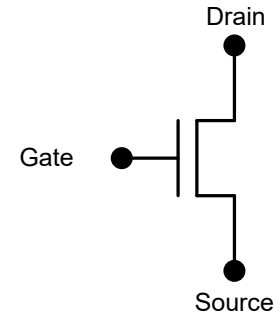


Area allocations shown to relative scale:

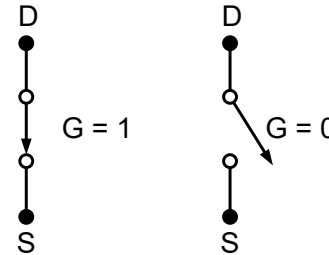
- Not enough information at this point to determine whether this very small capacitance can be neglected
- Will answer this important question later

Review from Last Time

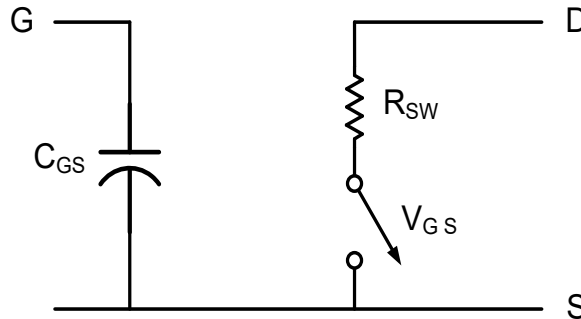
Model Summary (for n-channel)



1. Switch-Level model



2. Improved switch-level model

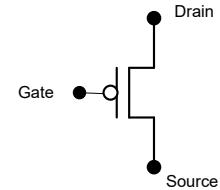


Switch closed for $V_{GS} = \text{large}$
Switch open for $V_{GS} = \text{small}$

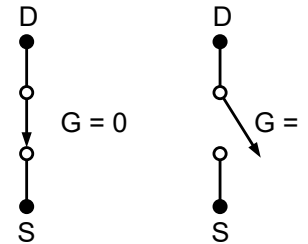
Other models will be developed later

Review from Last Time

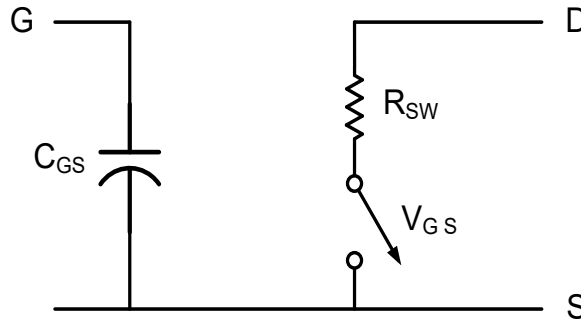
Model Summary (for p-channel)



1. Switch-Level model



2. Improved switch-level model



Switch closed for $|V_{GS}| = \text{large}$
Switch open for $|V_{GS}| = \text{small}$

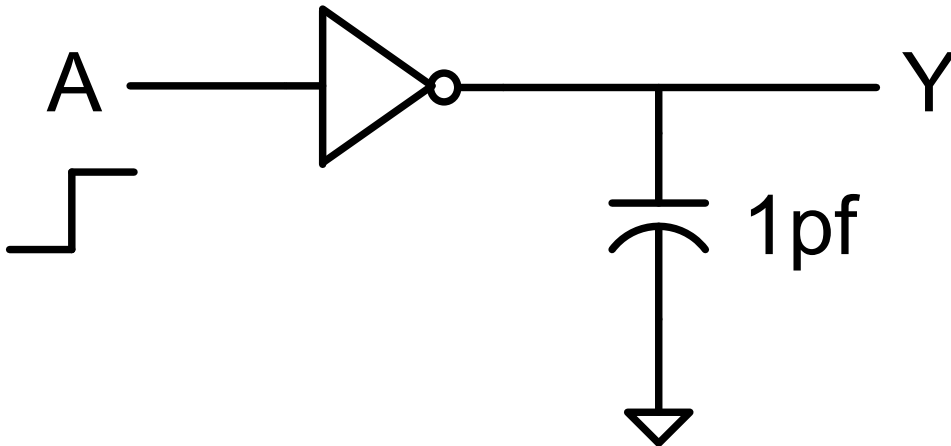
Other models will be developed later

Propagation Delay

Example

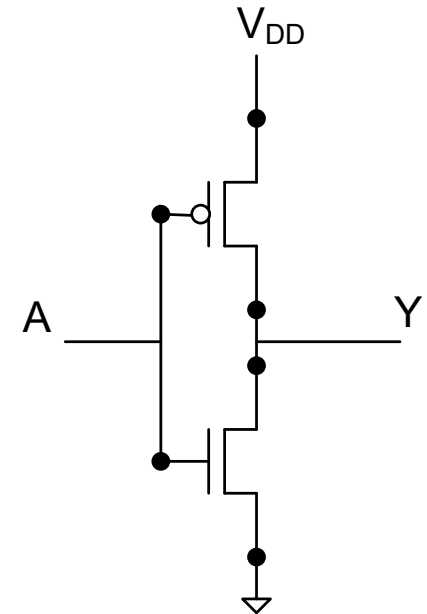
What are t_{HL} and t_{LH} ?

Assume $V_{DD}=5V$



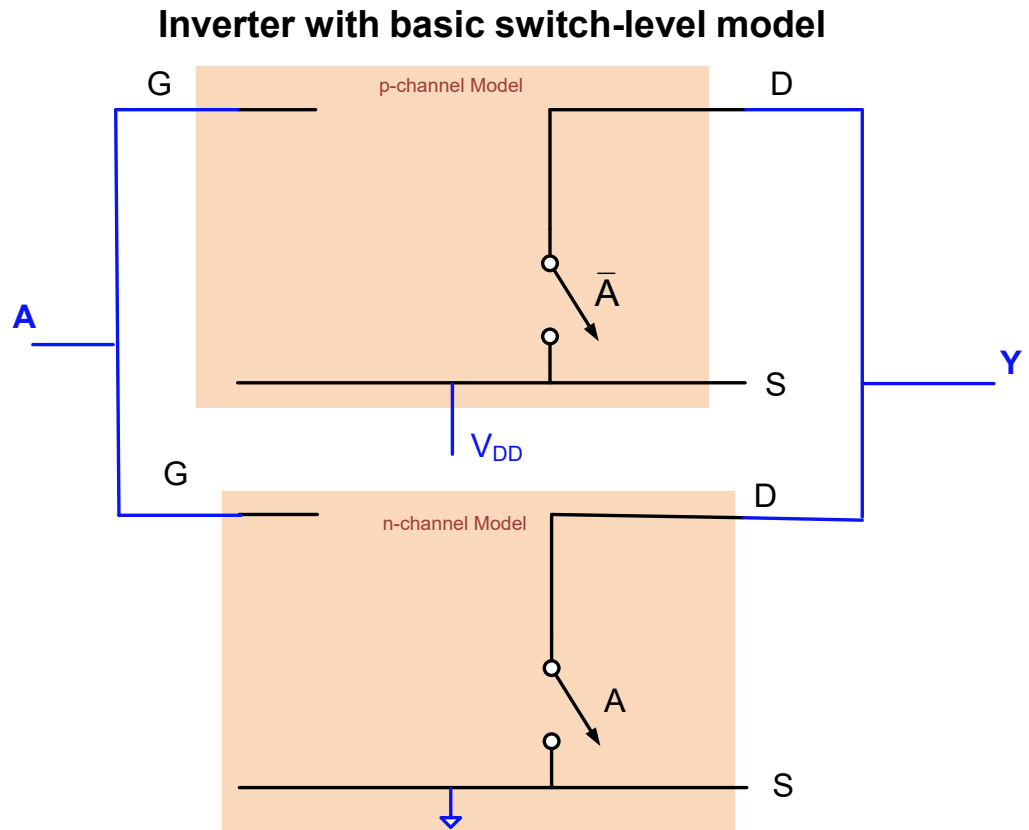
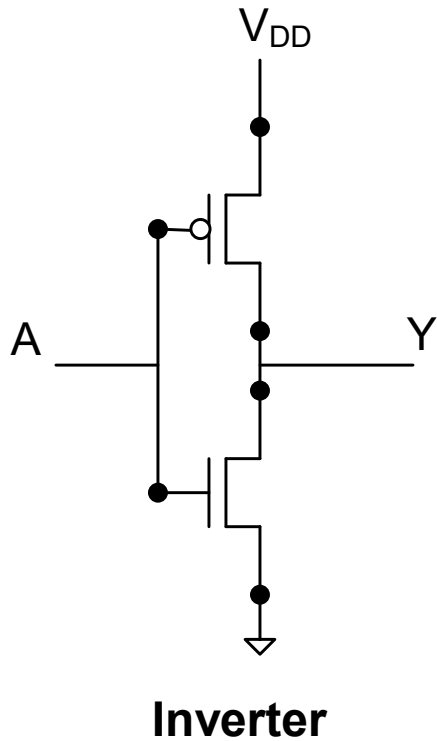
With basic switch level model ?

With improved switch level model ?

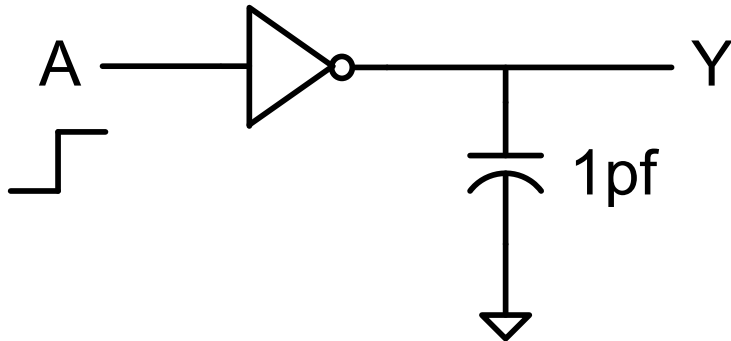


Inverter

Example

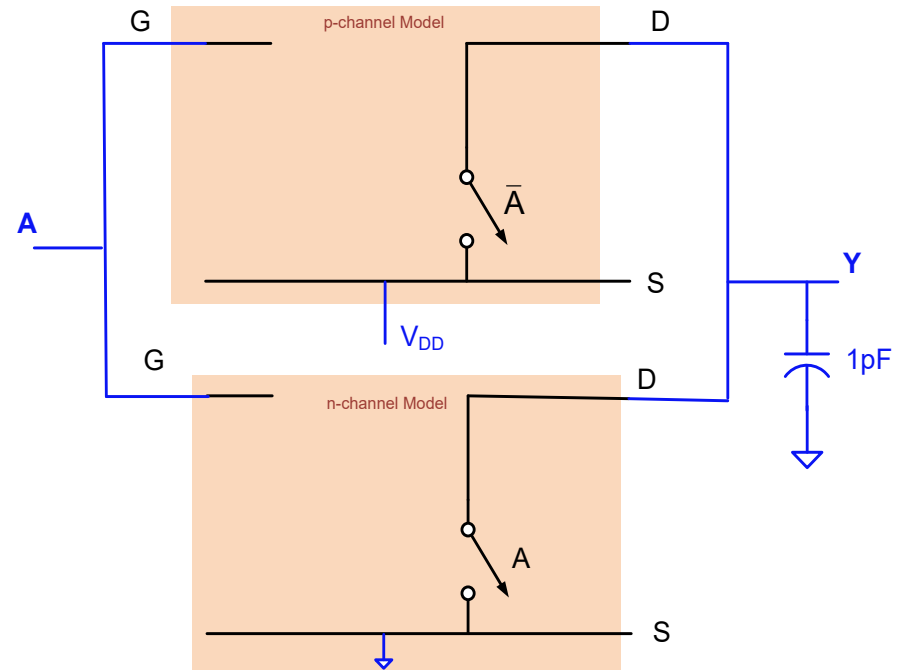
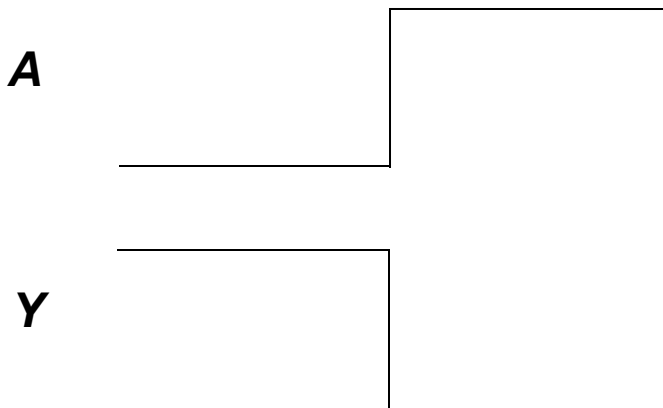


Example *What are t_{HL} and t_{LH} at output?*



Assume ideal step at A input

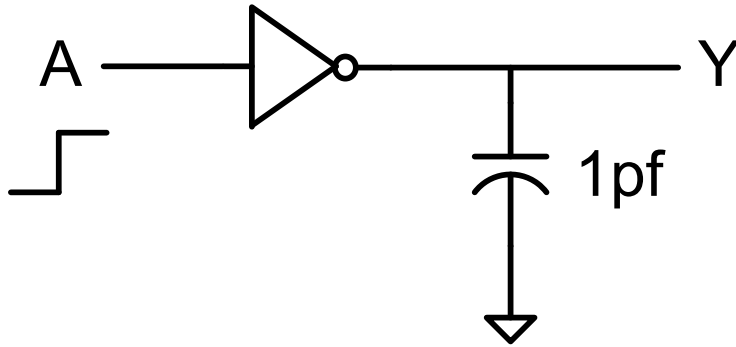
With basic switch level model



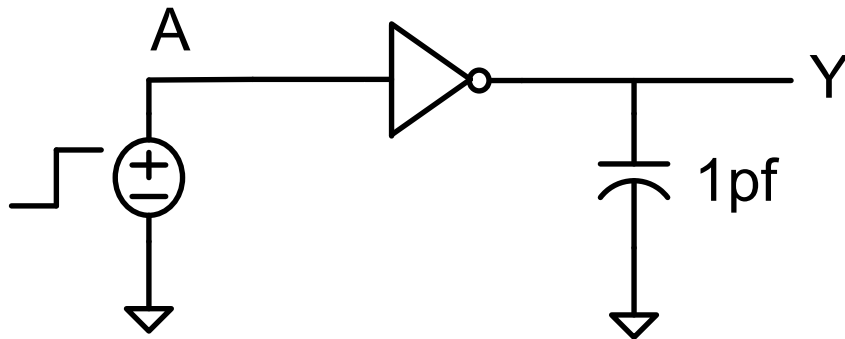
$$t_{HL} = t_{LH} = 0$$

Example (cont)

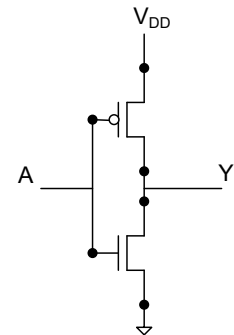
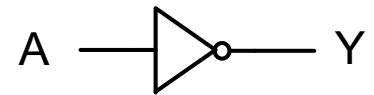
With simple switch-level model $t_{HL}=t_{LH}=0$



With improved model ?



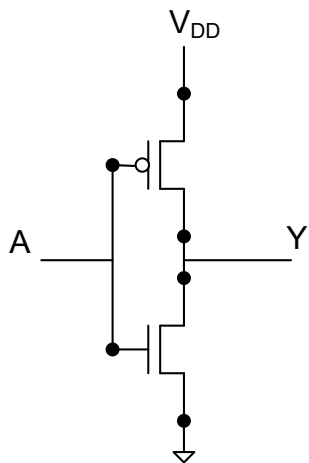
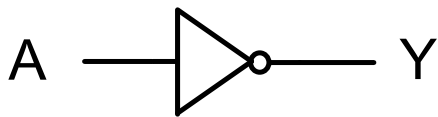
Inverter Model?



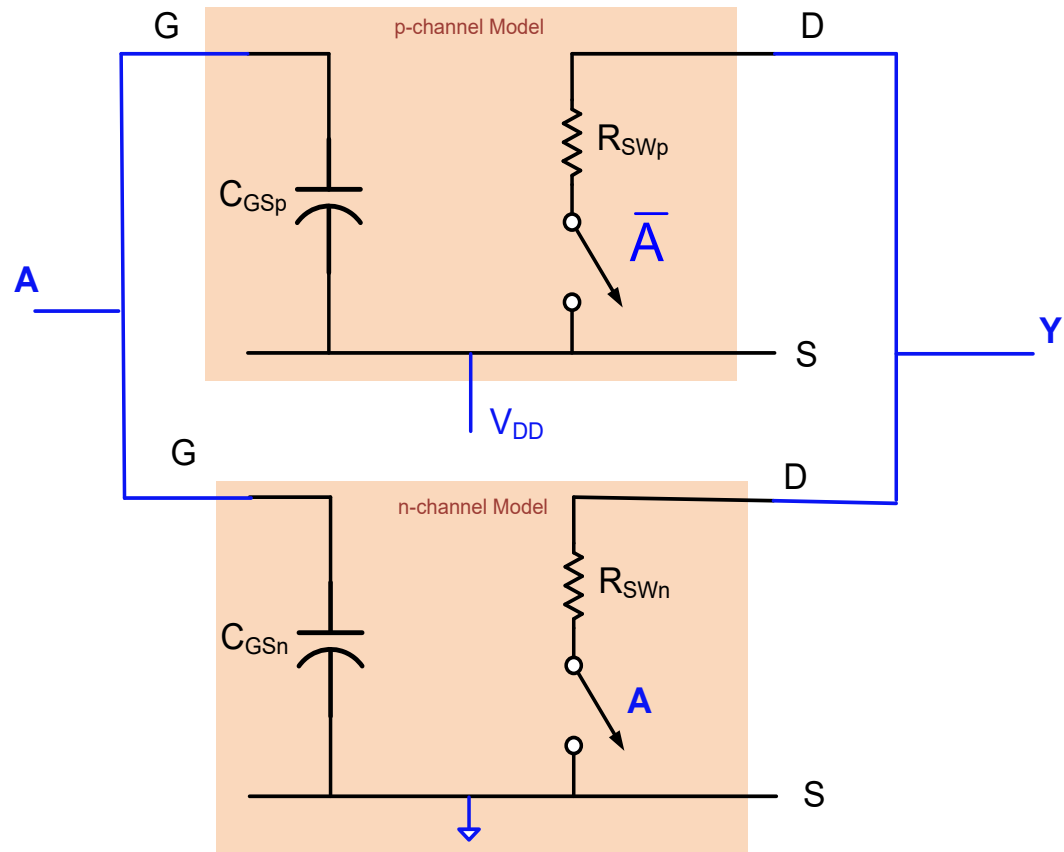
Example (cont)

Inverter with improved model

Inverter Model

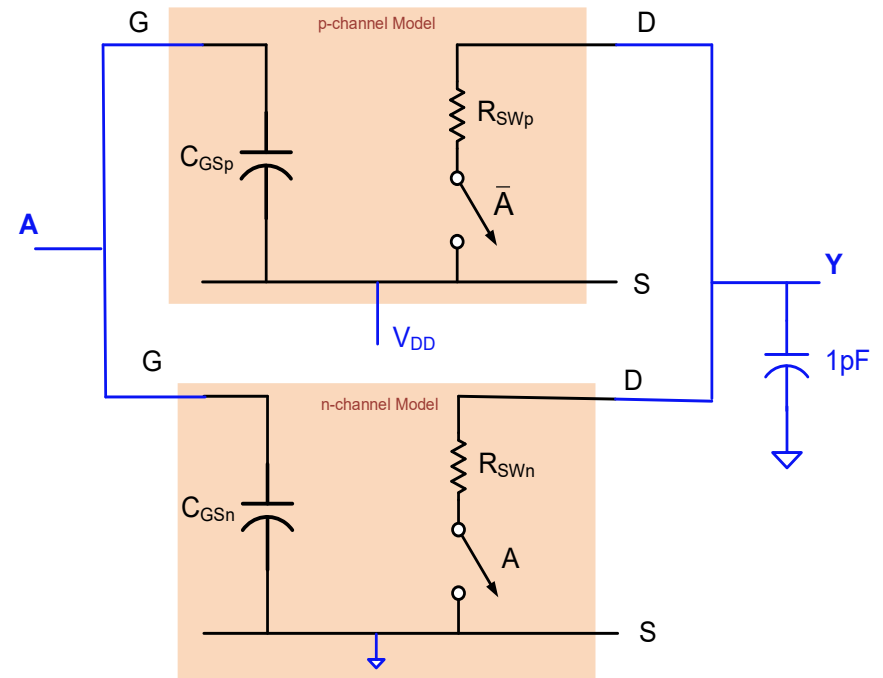
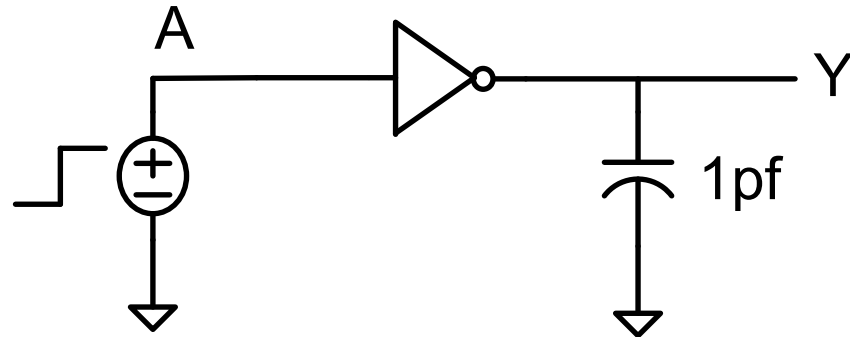


Inverter with Improved Model

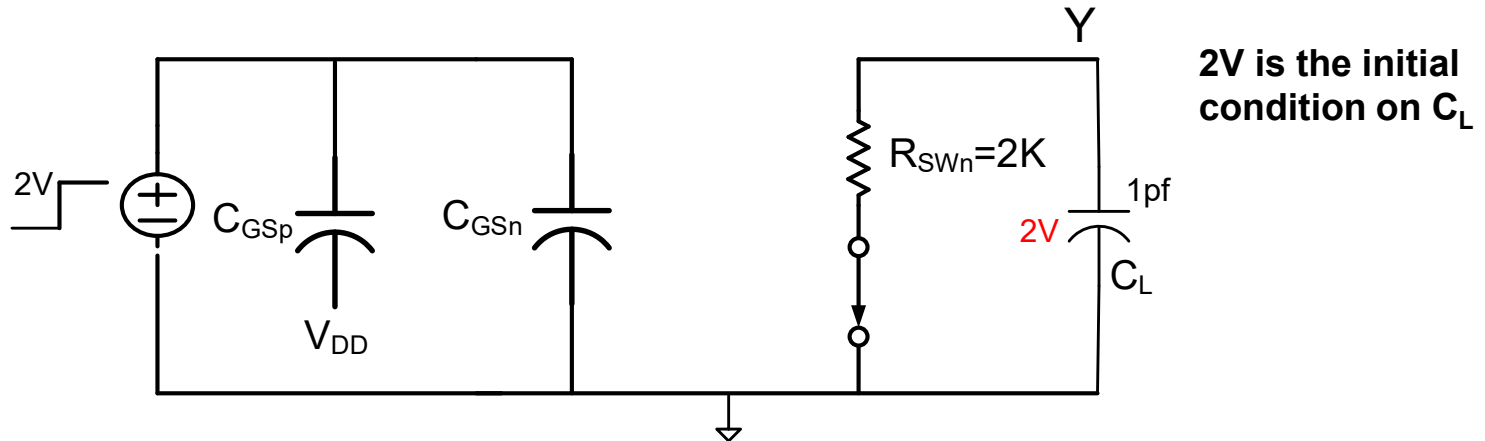


Example (cont)

With improved model $t_{HL}=?$



To initiate a HL output transition, assume Y has been in the high state for a long time, lower switch closes at time $t=0$, and upper switch opens just prior to $t=0$

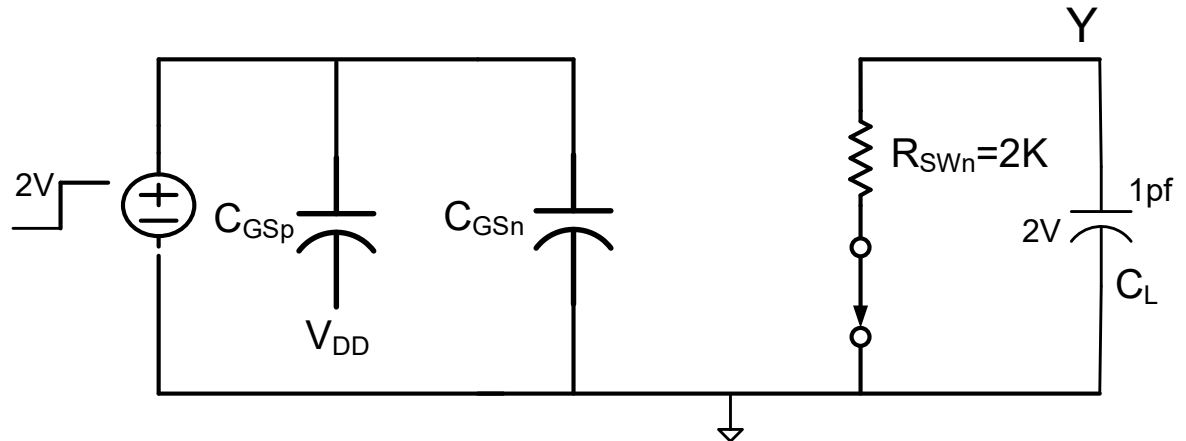


2V is the initial condition on C_L

Example (cont)

With improved model

$$t_{HL}=?$$



Recognize circuit (specifically on right) as a first-order RC network

Recall: Step response of any first-order network with LHP pole can be written as

$$y(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

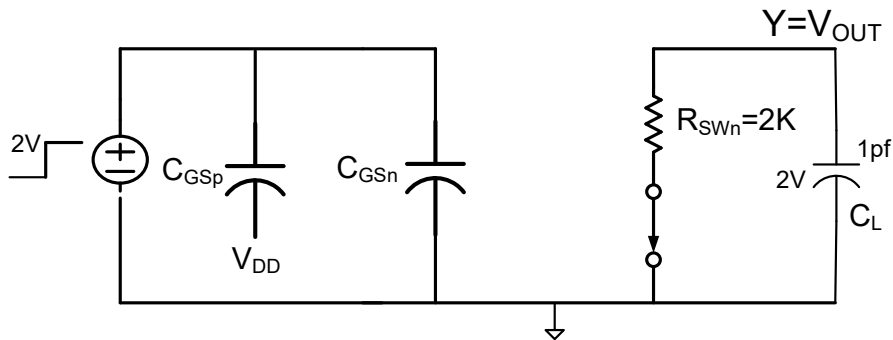
where F is the final value, I is the initial value and τ is the time constant of the circuit

(from Chapter 7 of Nilsson and Riedel)

For the circuit above, $F=0$, $I=2$ and $\tau = R_{SWn} C_L$

Example (cont)

With improved model

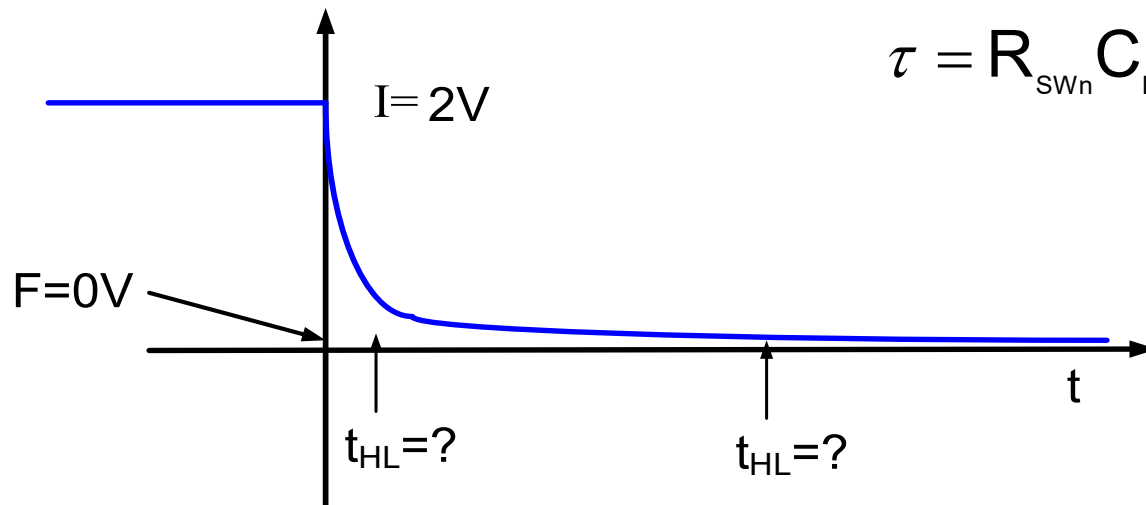


$$t_{HL} = ?$$

$$V_{OUT}(t) = F + (1 - F)e^{-\frac{t}{\tau}}$$

$$V_{OUT}(t) = 2e^{-\frac{t}{\tau}}$$

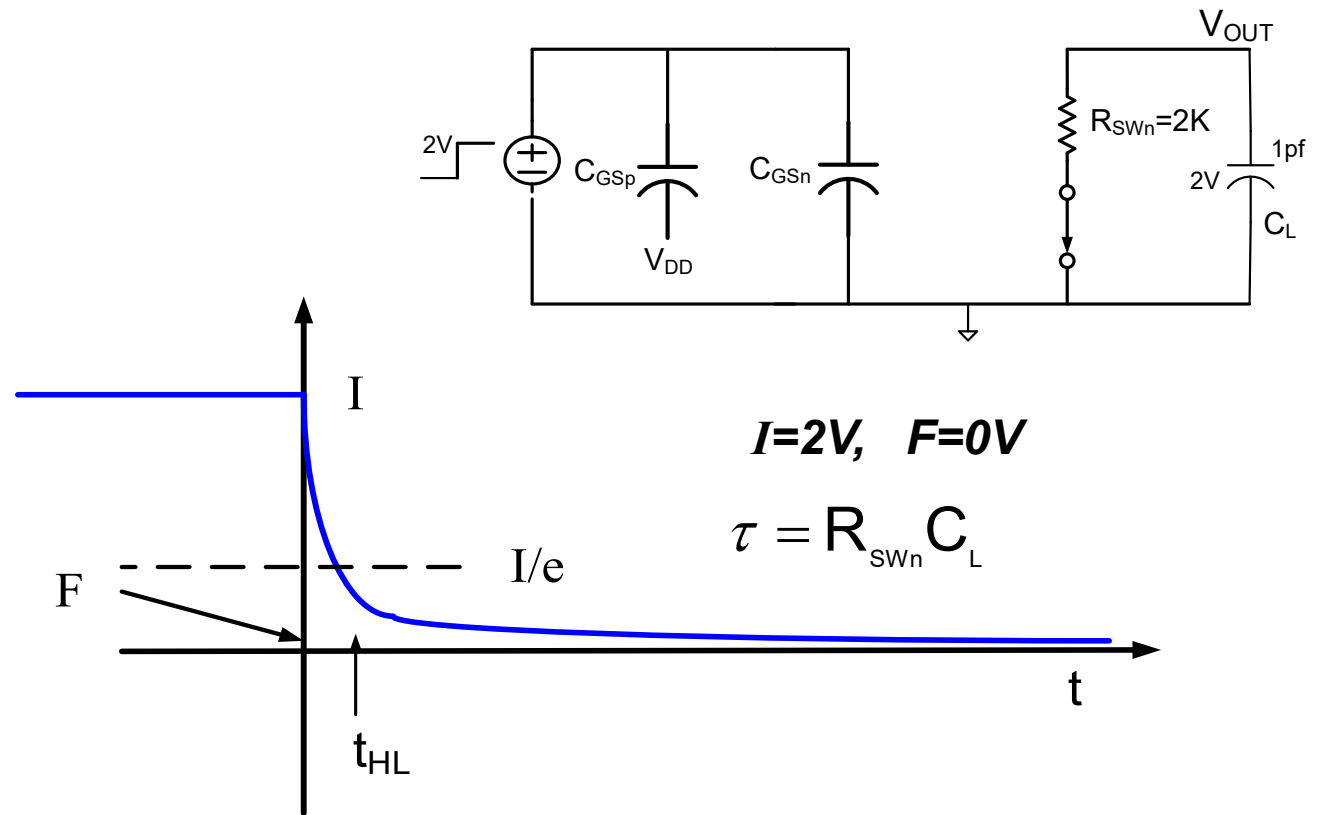
$$\tau = R_{SWn} C_L$$



how is t_{HL} defined?

Example (cont)

$t_{HL}=?$



Define t_{HL} to be the time taken for output to drop to I/e

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}} \quad \longrightarrow \quad \frac{I}{e} = F + (I - F)e^{-\frac{t_{HL}}{\tau}}$$

Is this simply a mathematical definition or does it have some practical significance?

t_{HL} as defined here and as verified by experimental verification has proven useful at analytically predicting response time of circuits

Example (cont)

With improved model

$$\frac{I}{e} = F + (I - F)e^{-\frac{t_{HL}}{\tau}}$$

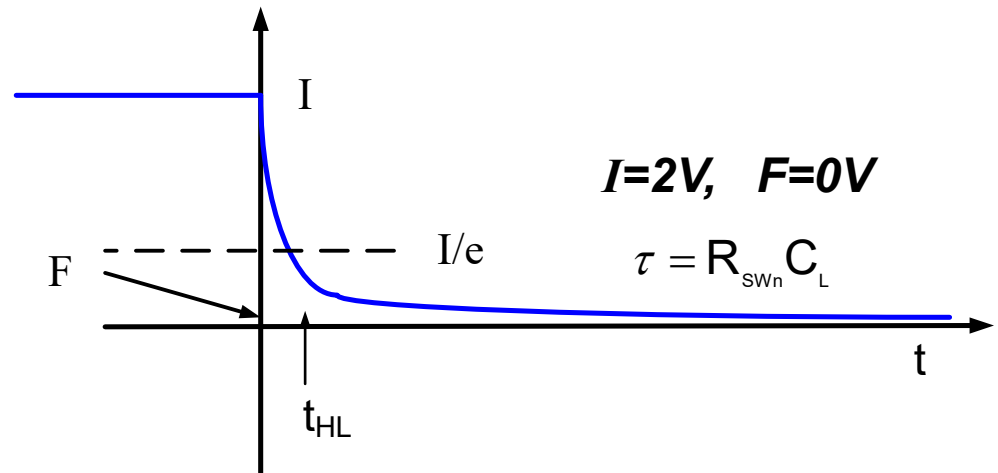
$$\frac{I}{e} = Ie^{-\frac{t_{HL}}{\tau}}$$

$$\frac{1}{e} = e^{-\frac{t_{HL}}{\tau}}$$

$$t_{HL} = \tau$$



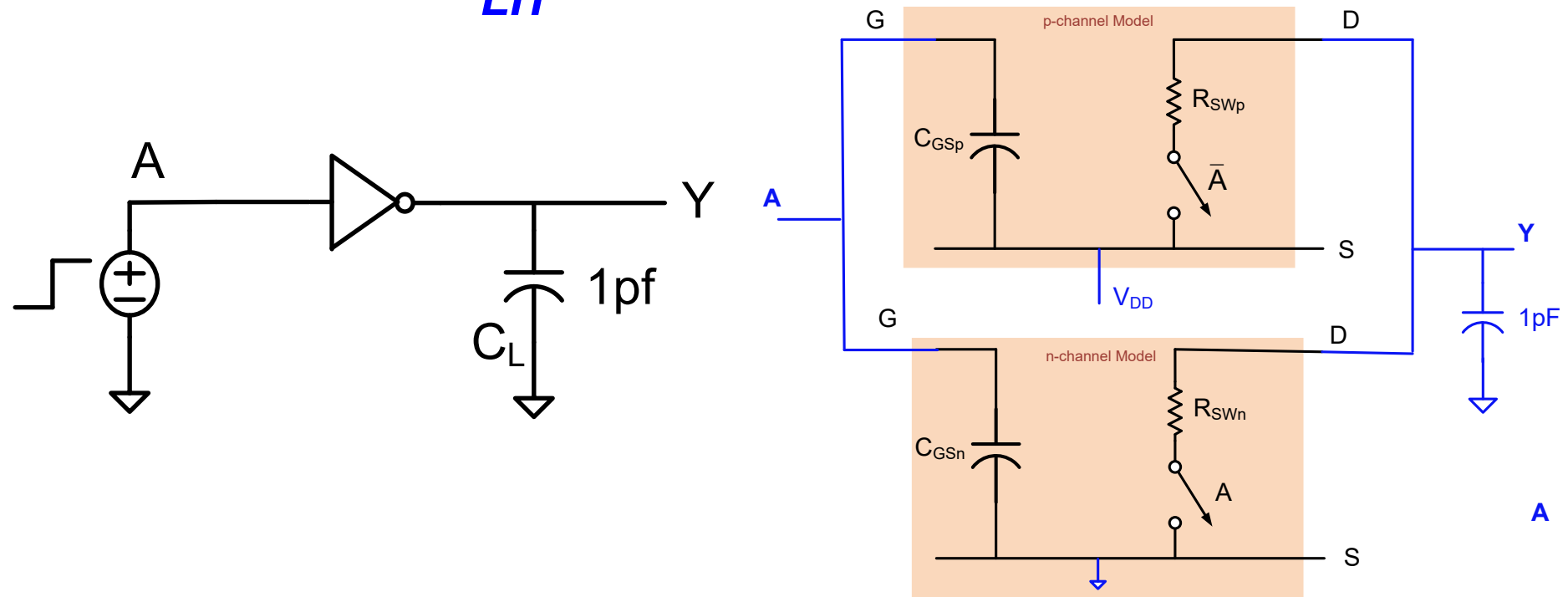
$$t_{HL} = R_{SWn} C_L$$



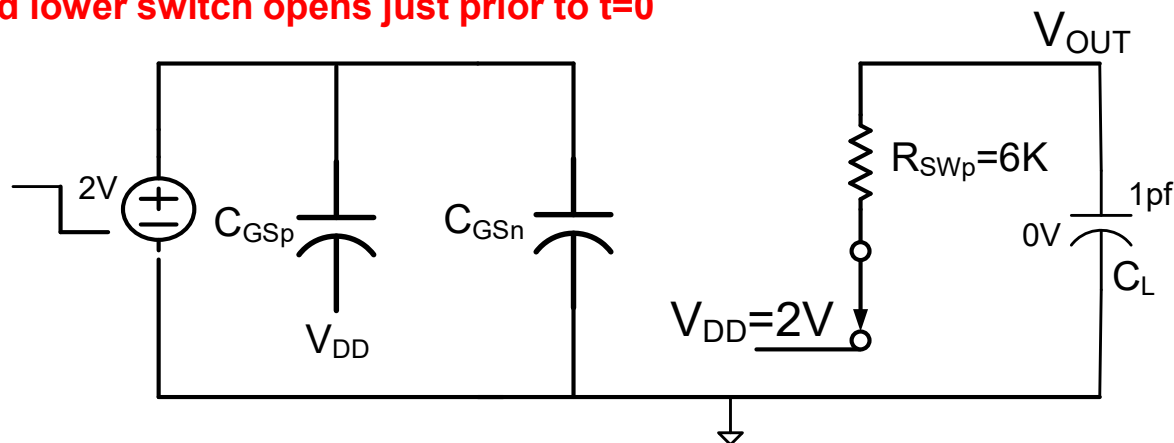
Both experimental results and accurate computer simulations show that this reasonably accurately predicts how quickly following stages recognize that a logic transition has taken place !!

Example (cont)

With improved model $t_{LH}=?$



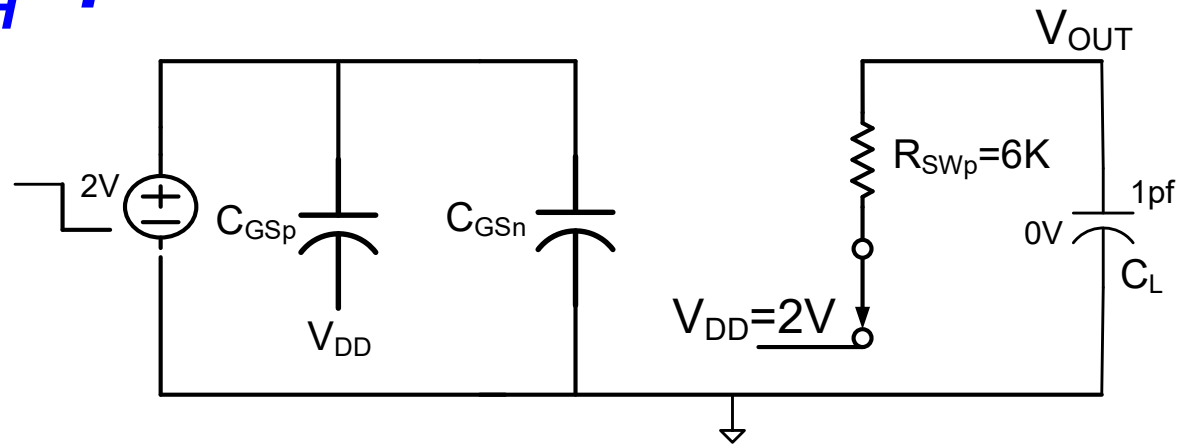
Assume output in low state for a long time and upper switch closes at time $t=0$ and lower switch opens just prior to $t=0$



0V is the initial condition on C_L

Example (cont)

With improved model $t_{LH}=?$

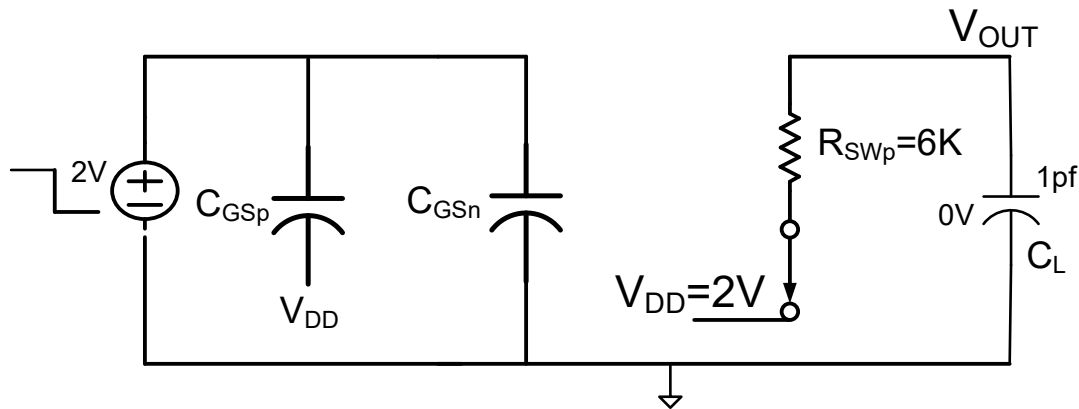


$$y(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

For this circuit (specifically on the right), $F=2$, $I=0$ and $\tau = R_{SWp} C_L$

Example (cont)

With improved model

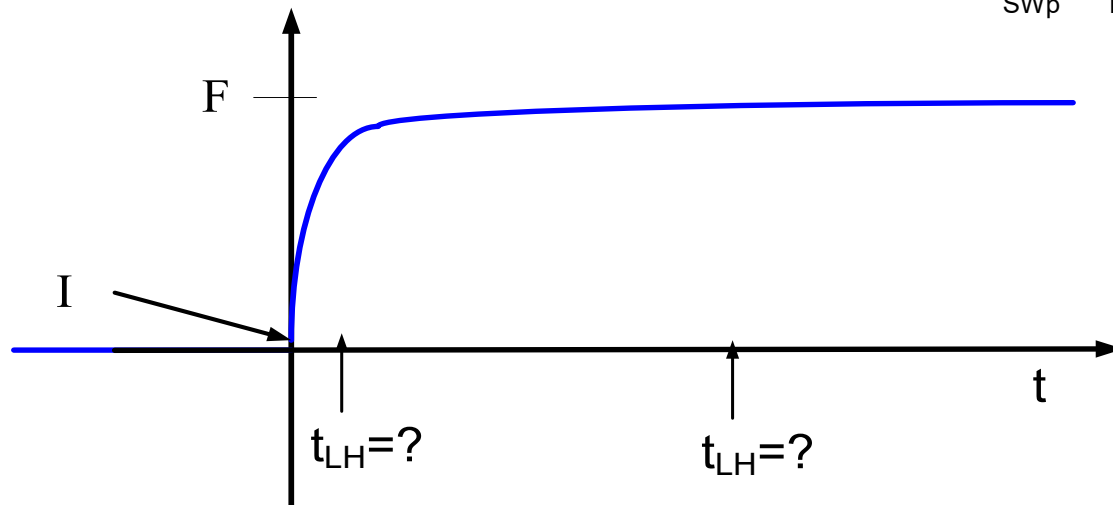


$t_{LH}=?$

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

$$V_{OUT}(t) = 2\left(1 - e^{-\frac{t}{\tau}}\right)$$

$$\tau = R_{SWp} C_L$$



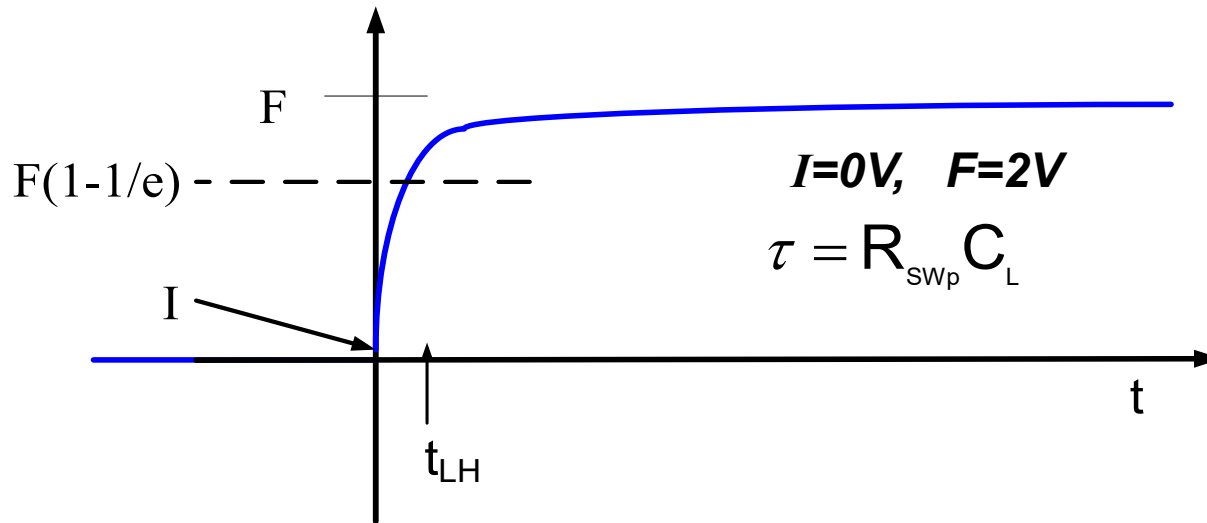
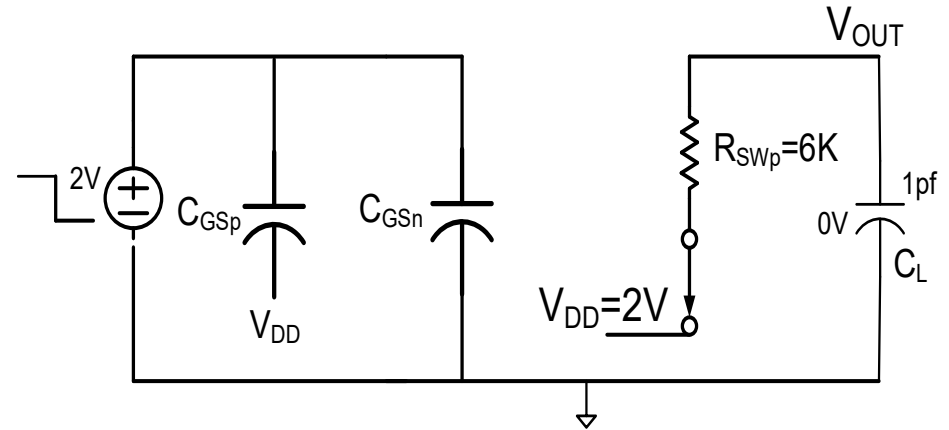
how is t_{LH} defined?

Example (cont)

With improved model

$$t_{LH}=?$$

Define t_{LH} as shown on figure



t_{LH} as defined has proven useful for analytically predicting response time of circuits

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}} \quad \longrightarrow \quad F\left(1 - \frac{1}{e}\right) = F + (I - F)e^{-\frac{t_{LH}}{\tau}}$$

Example (cont)

With improved model

$$t_{LH}=?$$

$$F\left(1 - \frac{1}{e}\right) = F + (I - F)e^{-\frac{t_{LH}}{\tau}}$$

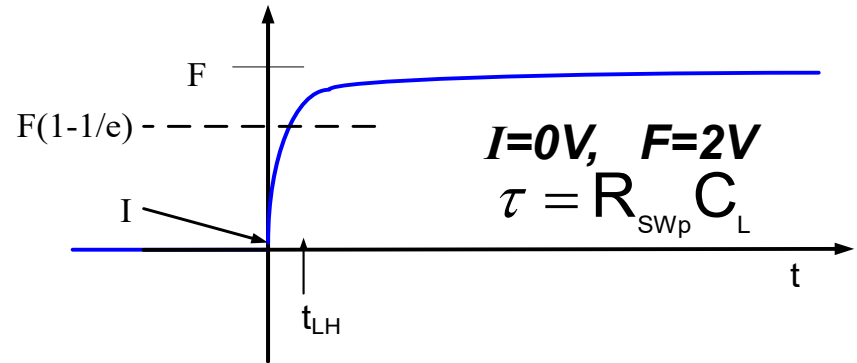
$$F\left(1 - \frac{1}{e}\right) = F + (F)e^{-\frac{t_{LH}}{\tau}}$$

$$1 - \frac{1}{e} = 1 + e^{-\frac{t_{LH}}{\tau}}$$

$$t_{LH} = \tau$$

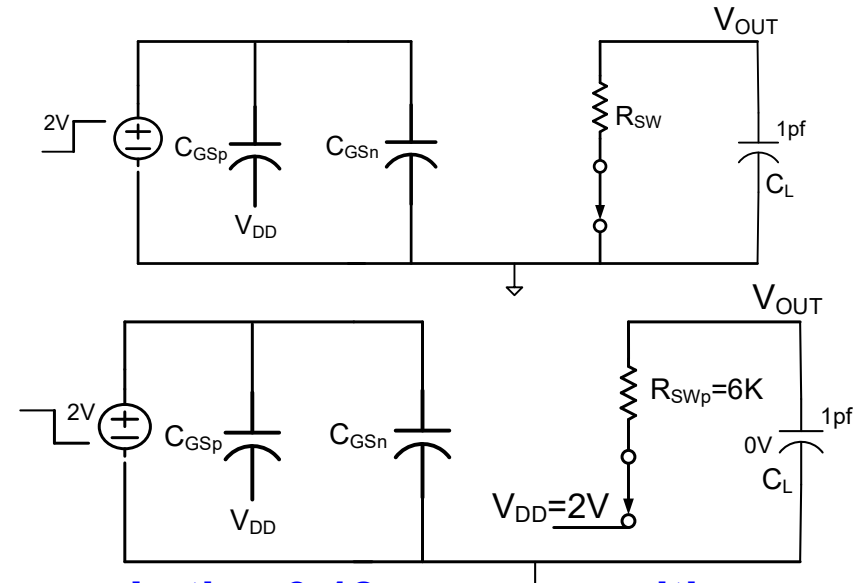


$$t_{LH} = R_{SWp} C_L$$



Example (cont)

With improved model



In the 0.18u process with minimum-sized devices

$$t_{HL} \cong R_{SWn} C_L$$

$$= 2K \bullet 1pF = 2n \text{ sec}$$

$$t_{LH} \cong R_{SWp} C_L$$

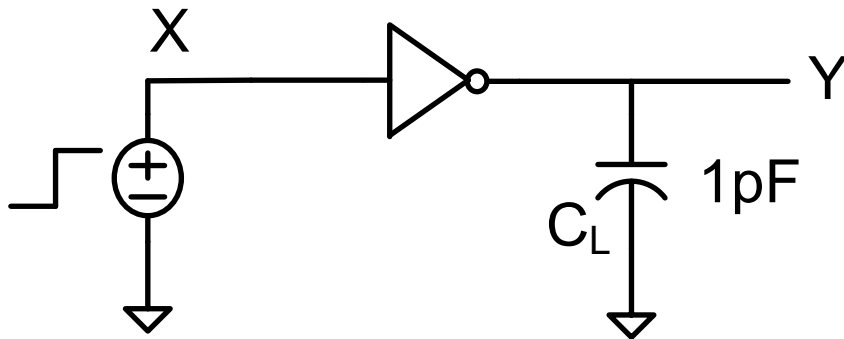
$$= 6K \bullet 1pF = 6n \text{ sec}$$

Note this circuit is quite fast !

Note that t_{HL} is much shorter than t_{LH}

Often C_L will be even smaller and the circuit will be much faster !!

Summary: What is the delay of a minimum-sized inverter driving a 1pF load?

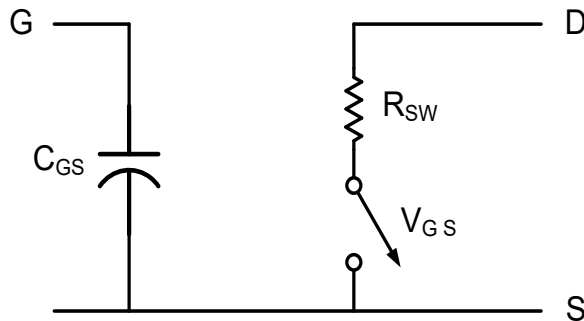
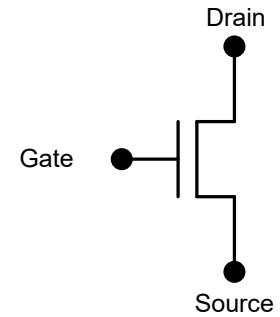


In a 0.18 μ process

$$t_{HL} \cong R_{SWn} C_L = 2K \bullet 1pF = 2n \text{ sec}$$

$$t_{LH} \cong R_{SWp} C_L = 6K \bullet 1pF = 6n \text{ sec}$$

Improved switch-level model



Switch closed for $V_{GS} = \text{large}$

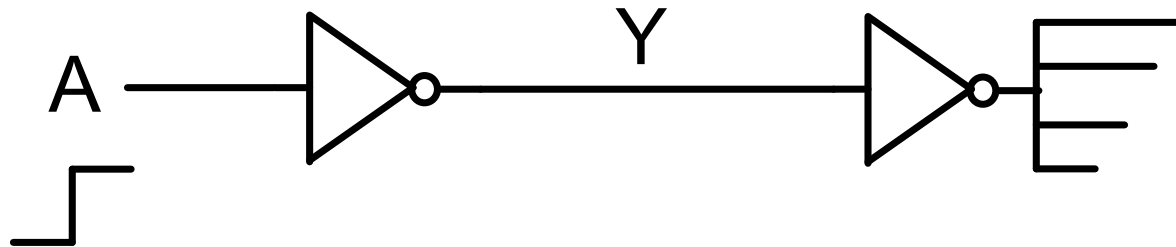
Switch open for $V_{GS} = \text{small}$

- Previous example showed why R_{SW} in the model was important
- But of what use is the C_{GS} which did not enter the previous calculations?

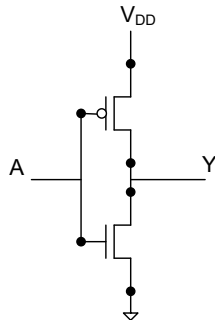
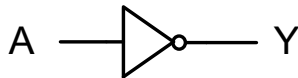
For minimum-sized devices in a 0.18μ process

$$C_{GS} \cong 1.5\text{fF} \quad R_{sw} \cong \left. \begin{array}{l} 2\text{K}\Omega \text{ n-channel} \\ 6\text{K}\Omega \text{ p-channel} \end{array} \right\}$$

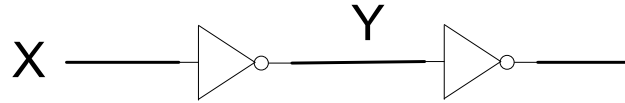
One gate often drives one or more other gates !



What are t_{HL} and t_{LH} ?



Example: What is the delay of a minimum-sized inverter driving another identical device?

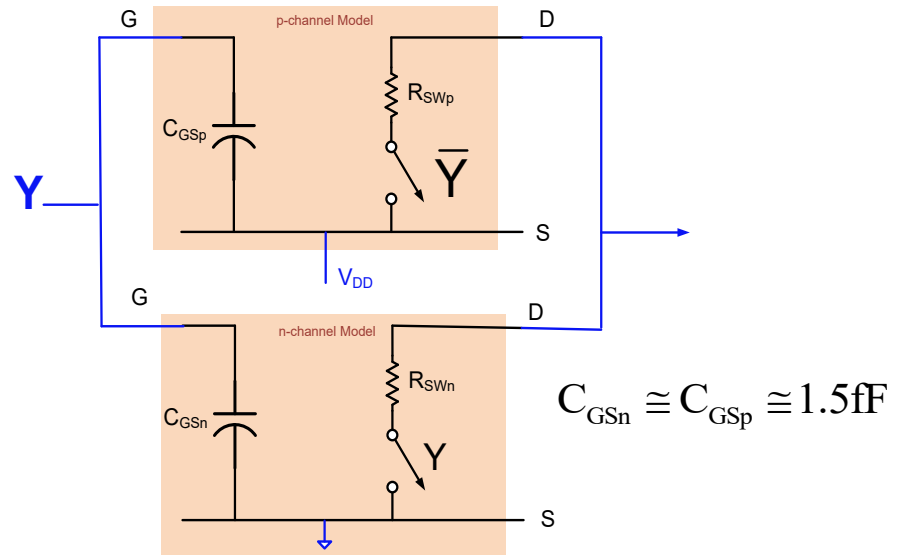


?

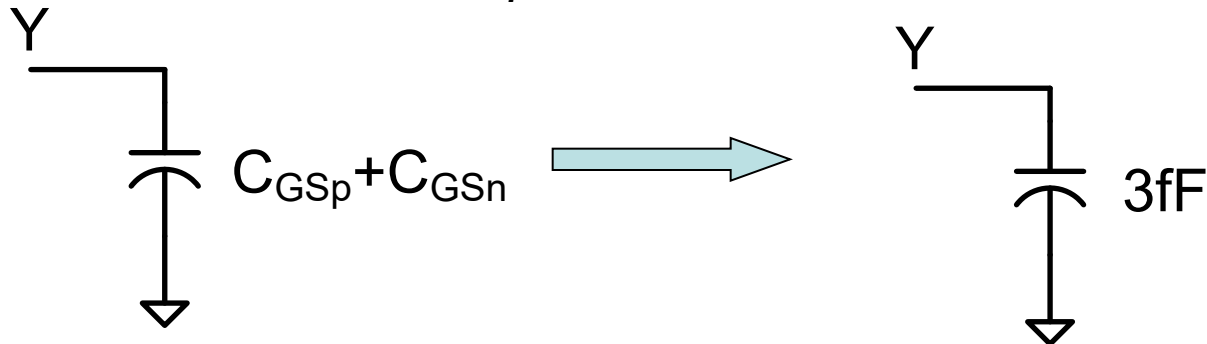


Load on first inverter

C_{GSn} and C_{GSp} both 1.5fF

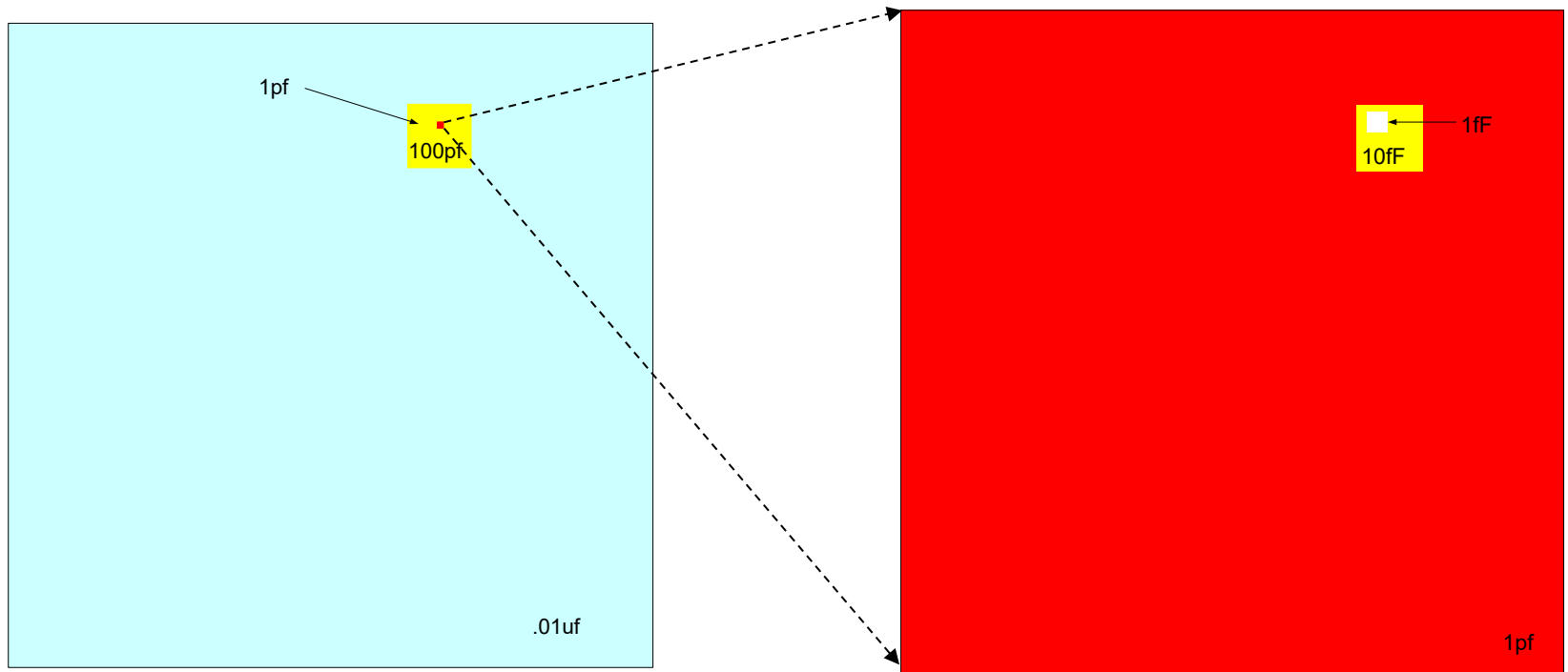


Loading effects same whether C_{GSp} and/or C_{GSn} connected to V_{DD} or GND



For convenience, will reference both to ground

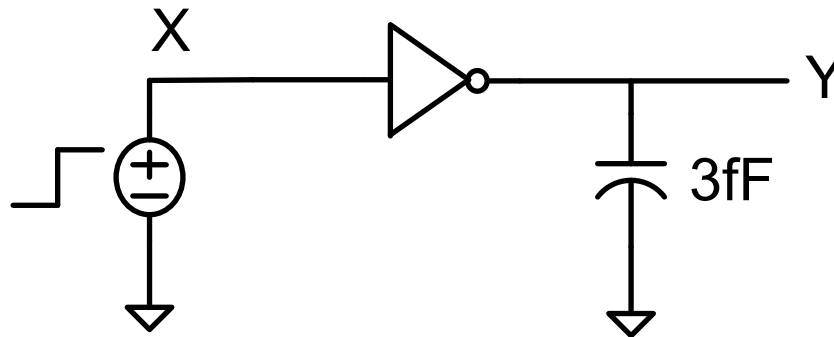
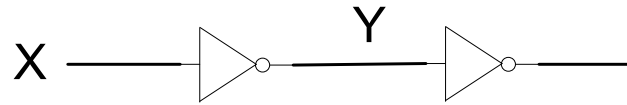
Is a capacitor of 1.5fF small enough to be neglected?



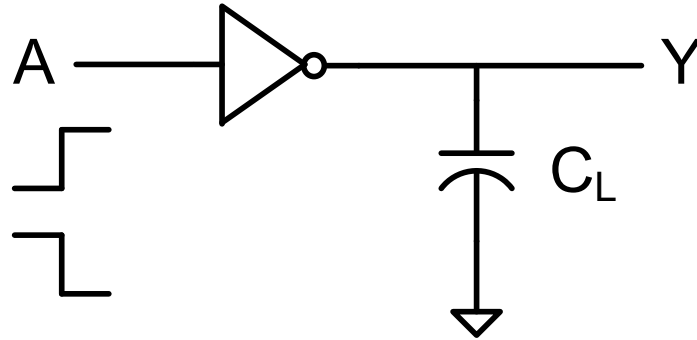
Area allocations shown to relative scale:

- This example will provide insight into the answer of the question

Example: What is the delay of a minimum-sized inverter driving another identical device? Assume $V_{DD}=2V$



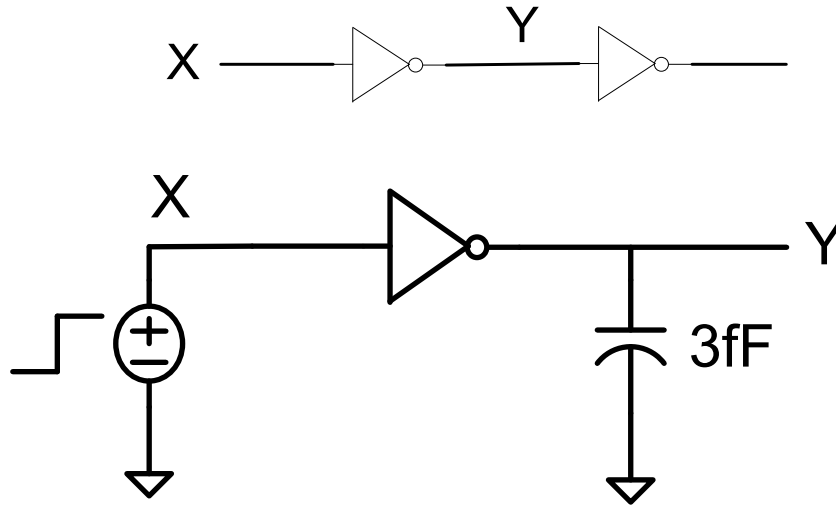
Generalizing the Previous Analysis to Arbitrary Load



$$t_{HL} \cong R_{SWn} C_L$$

$$t_{LH} \cong R_{SWp} C_L$$

Example: What is the delay of a minimum-sized inverter driving another identical device?



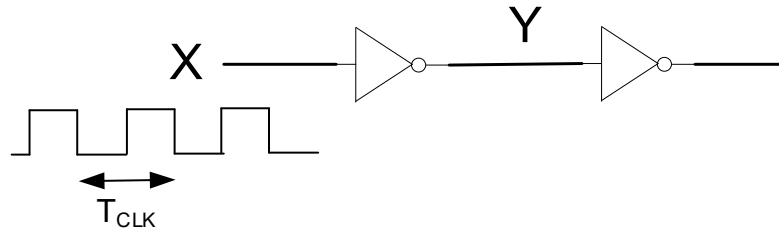
$$t_{HL} \cong R_{SWn} C_L = 2K \bullet 3fF = 6p \text{ sec}$$

$$t_{LH} \cong R_{SWp} C_L = 6K \bullet 3fF = 18p \text{ sec}$$

Do gates really operate this fast?

What would be the maximum clock rate for acceptable operation?

Example: What is the delay of a minimum-sized inverter driving another identical device?



$$t_{HL} \cong R_{SWn} C_L = 6p \text{ sec}$$

$$t_{LH} \cong R_{SWp} C_L = 18p \text{ sec}$$

What would be the maximum clock rate for acceptable operation?

$$T_{CLK-min} = t_{HL} + t_{LH}$$

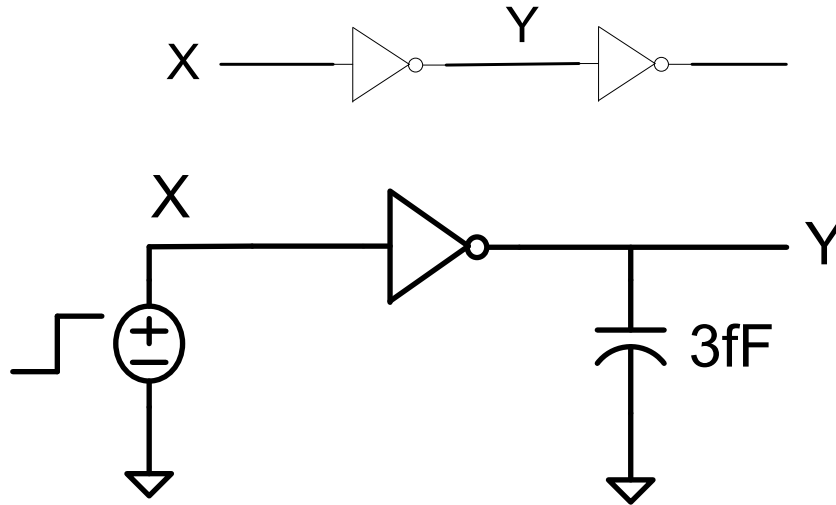
$$f_{CLK-max} = \frac{1}{T_{CLK-min}} = \frac{1}{24psec} = 40GHz$$

And much faster in a finer feature process !!

??????

What would be the implications of allowing for 10 levels of logic and 10 loads (FanOut=10)?

Example: What is the delay of a minimum-sized inverter driving another identical device? SUMMARY

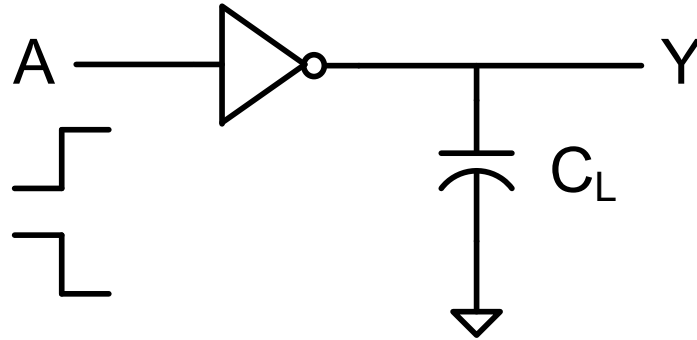


$$t_{HL} \cong R_{SWn} C_L = 2K \bullet 3fF = 6p \text{ sec}$$

$$t_{LH} \cong R_{SWp} C_L = 6K \bullet 3fF = 18p \text{ sec}$$

*This is very fast but even the small 1.5fF capacitors are not negligible !
These capacitors play a key role in determining the speed of a circuit !*

Response time of logic gates



$$t_{HL} \cong R_{SWn} C_L$$

$$t_{LH} \cong R_{SWp} C_L$$

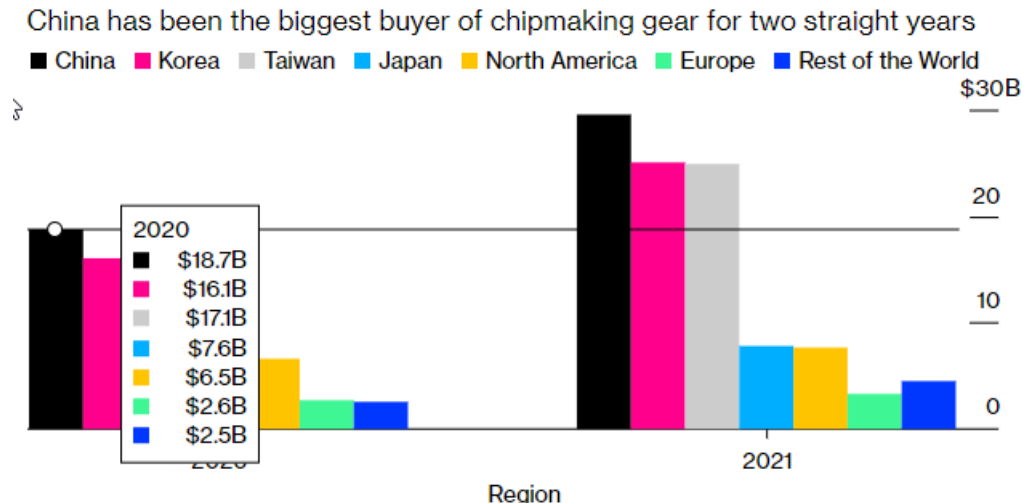
- Logic Circuits can operate very fast
- Extremely small parasitic capacitances play key role in speed of a circuit

Some Observations about Technology

Are the larger feature size technologies still used by industry today in the US or abroad?

GlobalData predicts that the Chinese market will play a much smaller role for foreign suppliers by 2030. More than 90% of the chips sold and used worldwide involve low-process production technology.

<https://www.investmentmonitor.ai/analysis/china-lead-global-semiconductor-growth-2030#:~:text=Global%20semiconductor%20industry%20revolves%20around,Samsung%20Electronics%20and%20SK%20Hynix.>



Some Observations about Technology

Sept 2022

<https://technode.com/2021/03/04/where-china-is-investing-in-semiconductors-in-charts/>

China is the world's largest consumer of semiconductors, and the lion's share of revenue from purchasing these chips go to foreign firms. China consumed \$143.4 billion worth of wafers in 2020, and just 5.9% of them were produced by companies headquartered in China.



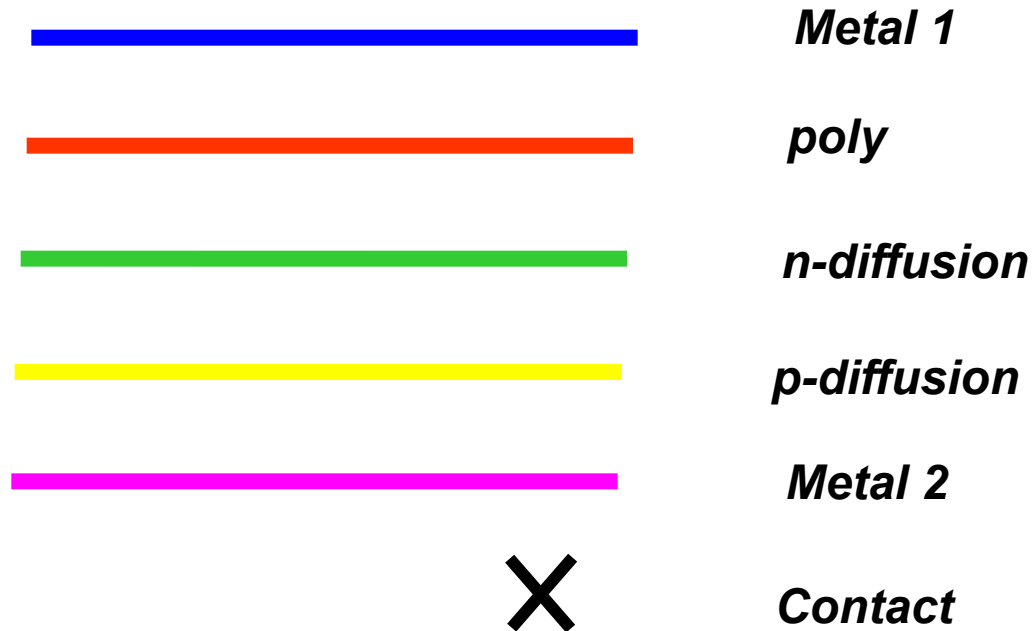
STATE OF THE U.S. SEMICONDUCTOR INDUSTRY

20 23

Stick Diagrams

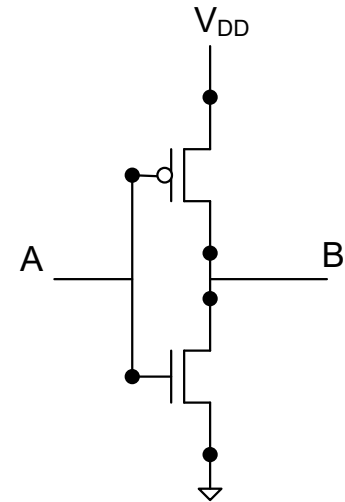
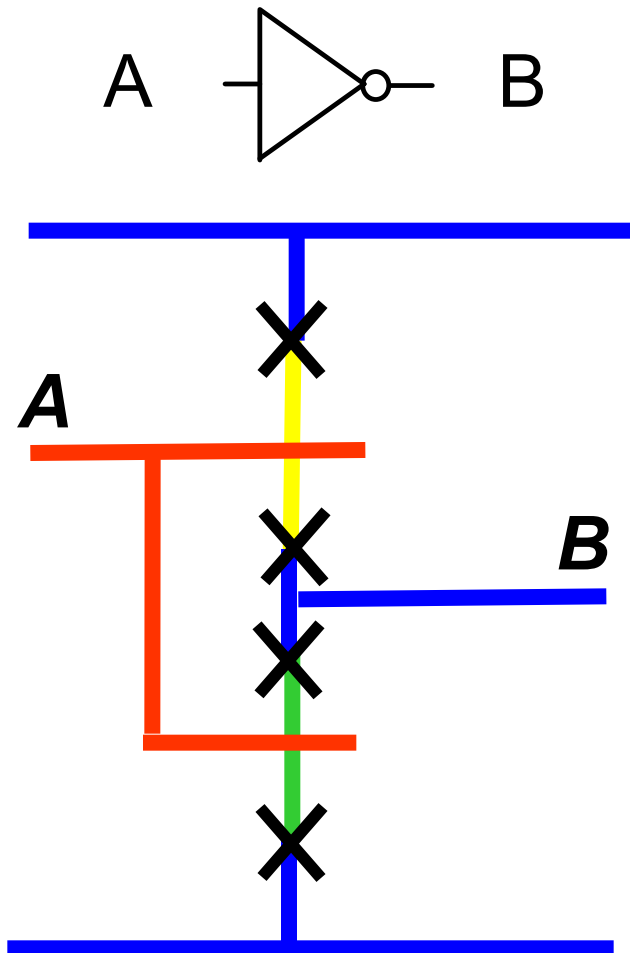
- It is often necessary to obtain information about placement, interconnect and physical-layer structure
- Stick diagrams are often used for small component-count blocks
- Approximate placement, routing, and area information can be obtained rather quickly with the use of stick diagrams

Stick Diagrams



Additional layers can be added and color conventions are personal

Stick Diagram

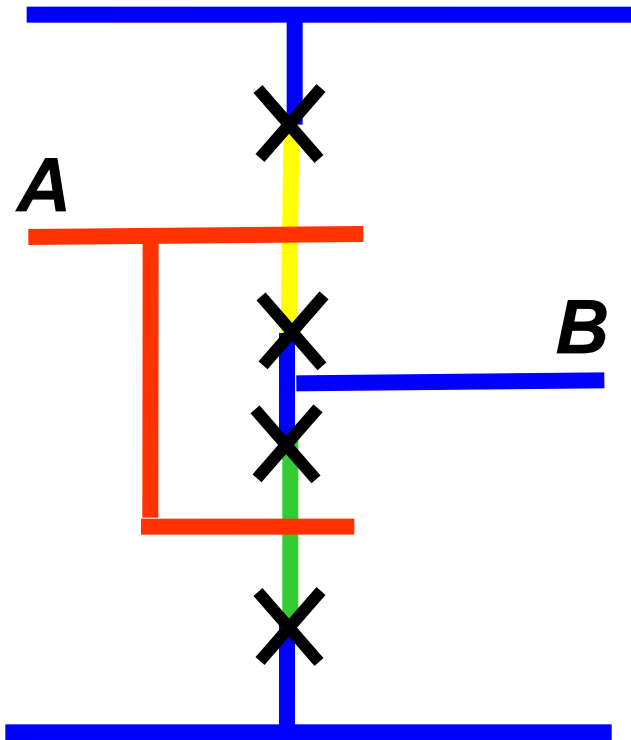
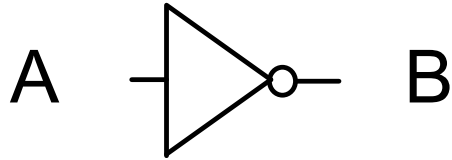


A stick diagram is not a layout but gives the basic structure (including location,, orientation and interconnects) that will be instantiated in the actual layout itself

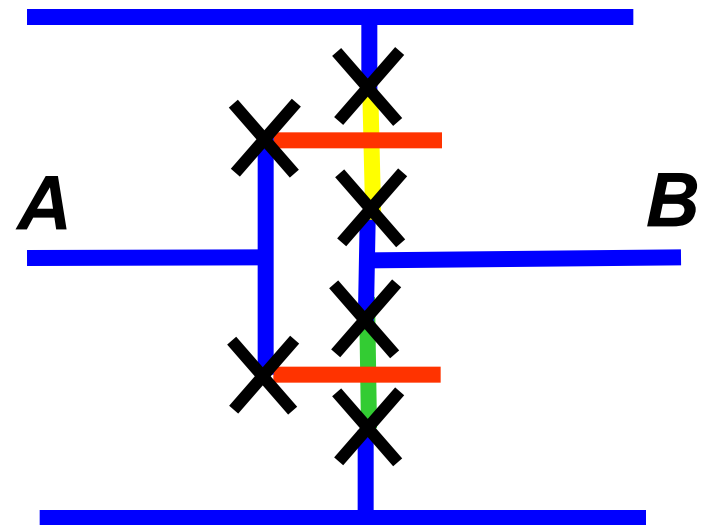
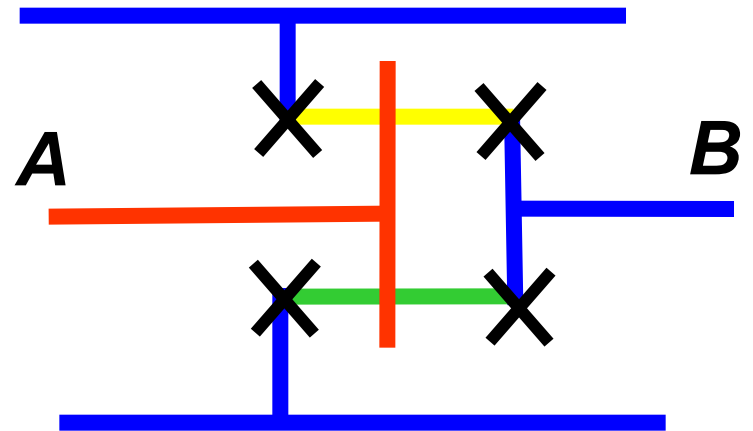
Modifications can be made much more quickly on a stick diagram than on a layout

Iteration may be needed to come up with a good layout structure

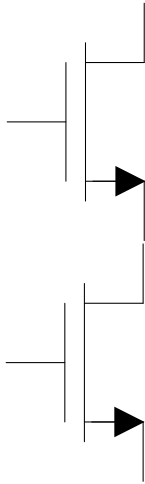
Stick Diagram



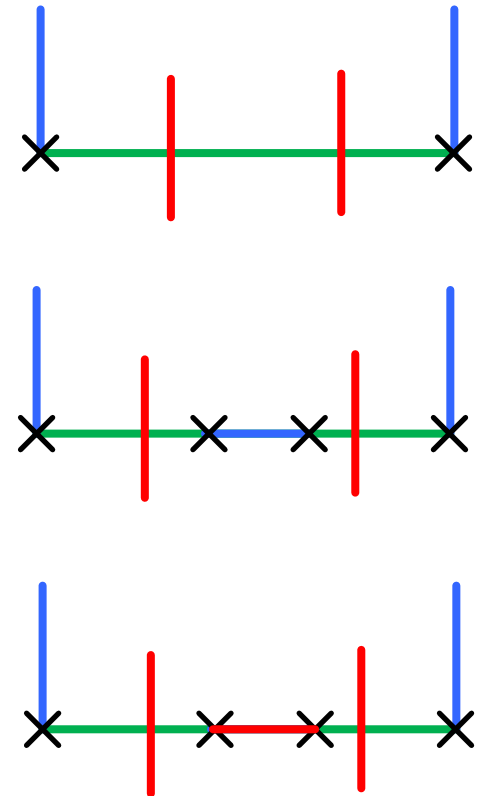
Alternate Representations



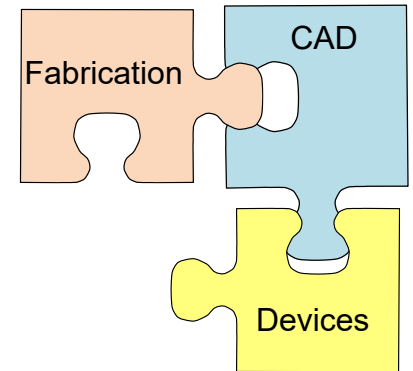
Stick Diagram



- **Source and drain notation suppressed**
- **Shared active can serve as interconnect**
- **No contact needed to shared active**
- **Multiple ways to layout even simple circuits**



Technology Files



- Provide Information About Process
 - Design Rules
 - Process Flow (Fabrication Technology)
 - Model Parameters
- Serve as Interface Between Design Engineer and Process Engineer
- Insist on getting information that is deemed important for a design
 - Limited information available in academia
 - Foundries often sensitive to who gets access to information
 - Customer success and satisfaction is critical to foundries

Technology Files

- Design Rules
- Process Flow (Fabrication Technology) (will discuss next)
- Model Parameters (will discuss in substantially more detail after device operation and more advanced models are introduced)

First – A preview of what the technology files look like !

Typical Design Rules

TABLE 2B.2
Design rules for a typical p-well CMOS process
 (See Table 2B.3 in color plates for graphical interpretation)

		Dimensions	
		Microns	Scalable
1.	p-well (CIF Brown, Mask #1 ^a)		
1.1	Width	5	4 λ
1.2	Spacing (different potential)	15	10 λ
1.3	Spacing (same potential)	9	6 λ
2.	Active (CIF Green, Mask #2)		
2.1	Width	4	2 λ
2.2	Spacing	4	2 λ
2.3	p ⁺ active in n-subs to p-well edge	8	6 λ
2.4	n ⁺ active in n-subs to p-well edge	7	5 λ
2.5	n ⁺ active in p-well to p-well edge	4	2 λ
2.6	p ⁺ active in p-well to p-well edge	1	λ
3.	Poly (POLY I) (CIF Red, Mask #3)		
3.1	Width	3	2 λ
3.2	Spacing	3	2 λ
3.3	Field poly to active	2	λ
3.4	Poly overlap of active	3	2 λ
3.5	Active overlap of poly	4	2 λ
4.	p ⁺ select (CIF Orange, Mask #4)		
4.1	Overlap of active	2	λ
4.2	Space to n ⁺ active	2	λ
4.3	Overlap of channel ^b	3.5	2 λ
4.4	Space to channel ^b	3.5	2 λ
4.5	Space to p ⁺ select	3	2 λ
4.6	Width	3	2 λ

Typical Design Rules (cont)

5.	Contact ^c (CIF Purple, Mask #6)		
5.1	Square contact, exactly	3×3	$2\lambda \times 2\lambda$
5.2	Rectangular contact, exactly	3×8	$2\lambda \times 6\lambda$
5.3	Space to different contact	3	2λ
5.4	Poly overlap of contact	2	λ
5.5	Poly overlap in direction of metal 1	2.5	2λ
5.6	Space to channel	3	2λ
5.7	Metal 1 overlap of contact	2	λ
5.8	Active overlap of contact	2	λ
5.9	p ⁺ select overlap of contact	3	2λ
5.10	Subs./well shorting contact, exactly	3×8	$2\lambda \times 6\lambda$
6.	Metal 1 ^d (CIF Blue, Mask #7)		
6.1	Width	3	2λ
6.2	Spacing	4	3λ
6.3	Maximum current density	0.8 mA/ μ	0.8 mA/ μ

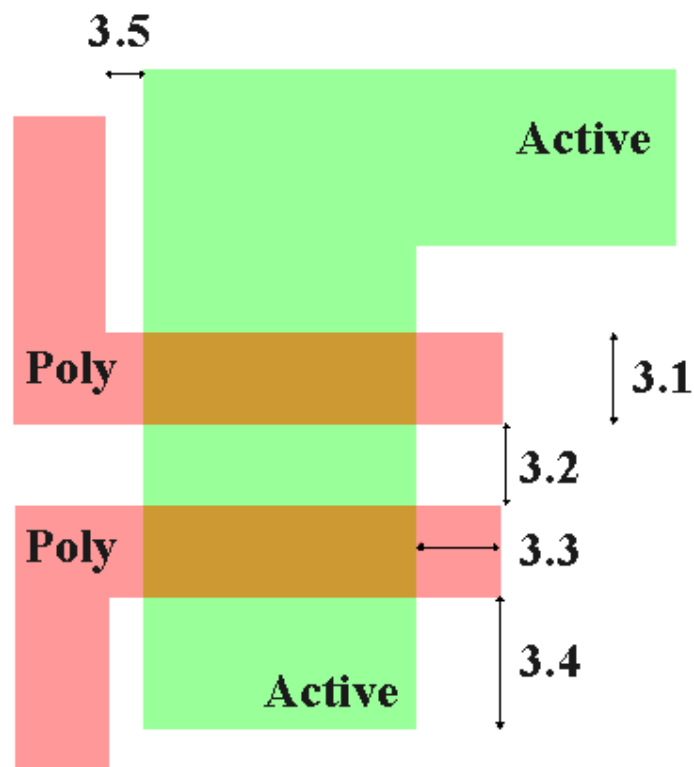
Typical Design Rules (cont)

7. Via ^e (CIF Purple Hatched, Mask #C1)		
7.1	Size, exactly	3×3 $2\lambda \times 2\lambda$
7.2	Separation	3 2λ
7.3	Space to poly edge	4 2λ
7.4	Space to contact	3 2λ
7.5	Overlap by metal 1	2 λ
7.6	Overlap by metal 2	2 λ
7.7	Space to active edge	3 2λ
8. Metal 2 (CIF Orange Hatched, Mask #C2)		
8.1	Width	5 3λ
8.2	Spacing	5 3λ
8.3	Bonding pad size	100×100 $100 \mu \times 100 \mu$
8.4	Probe pad size	75×75 $75 \mu \times 75 \mu$
8.5	Bonding pad separation	50 50μ
8.6	Bonding to probe pad	30 30μ
8.7	Probe pad separation	30 30μ
8.8	Pad to circuitry	40 40μ
8.9	Maximum current density	$0.8 \text{ mA}/\mu$ $0.8 \text{ mA}/\mu$
9. Passivation ^f (CIF Purple Dashed, Mask #8)		
9.1	Bonding pad opening	90×90 $90 \mu \times 90 \mu$
9.2	Probe pad opening	65×65 $65 \mu \times 65 \mu$
10. Metal 2 crossing coincident metal 1 and poly ^g		
10.1	Metal 1 to poly edge spacing when crossing metal 2	2 λ
10.2	Rule domain	2 λ
11. Electrode (POLY II) ^h (CIF Purple Hatched, Mask #A1)		
11.1	Width	3 2λ
11.2	Spacing	3 2λ
11.3	POLY I overlap of POLY II	2 λ
11.4	Space to contact	3 2λ

Typical Design Rules (cont)

SCMOS Layout Rules - Poly

Rule	Description	Lambda		
		SCMOS	SUBM	DEEP
3.1	Minimum width	2	2	2
3.2	Minimum spacing over field	2	3	3
3.2.a	Minimum spacing over active	2	3	4
3.3	Minimum gate extension of active	2	2	2.5
3.4	Minimum active extension of poly	3	3	4
3.5	Minimum field poly to active	1	1	1



Typical Process Description

Process scenario of major process steps in typical p-well CMOS process^a

1. Clean wafer
2. GROW THIN OXIDE
3. Apply photoresist
4. PATTERN P-WELL (MASK #1)
5. Develop photoresist
6. Deposit and diffuse p-type impurities
7. Strip photoresist
8. Strip thin oxide
9. Grow thin oxide
10. Apply layer of Si_3N_4
11. Apply photoresist
12. PATTERN Si_3N_4 (active area definition) (MASK #2)
13. Develop photoresist
14. Etch Si_3N_4
15. Strip photoresist
- Optional field threshold voltage adjust*
- A.1 Apply photoresist
- A.2 PATTERN ANTIMOAT IN SUBSTRATE (MASK #A1)
- A.3 Develop photoresist
- A.4 FIELD IMPLANT (n-type)
- A.5 Strip photoresist
16. GROW FIELD OXIDE
17. Strip Si_3N_4
18. Strip thin oxide
19. GROW GATE OXIDE
20. POLYSILICON DEPOSITION (POLY I)
21. Apply photoresist
22. PATTERN POLYSILICON (MASK #3)
23. Develop photoresist
24. ETCH POLYSILICON

Typical Process Description (cont)

- | | | |
|-------|--------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|------------|
| 25. | Strip photoresist
<i>Optional steps for double polysilicon process</i>
B.1 Strip thin oxide
B.2 GROW THIN OXIDE
B.3 POLYSILICON DEPOSITION (POLY II)
B.4 Apply photoresist
B.5 PATTERN POLYSILICON
B.6 Develop photoresist
B.7 ETCH POLYSILICON
B.8 Strip photoresist
B.9 Strip thin oxide | (MASK #B1) |
| <hr/> | | |
| 26. | Apply photoresist | |
| 27. | PATTERN P-CHANNEL DRAINS AND SOURCES AND
P ⁺ GUARD RINGS (p-well ohmic contacts) | (MASK #4) |
| 28. | Develop photoresist | |
| 29. | p ⁺ IMPLANT | |
| 30. | Strip photoresist | |
| 31. | Apply photoresist | |
| 32. | PATTERN N-CHANNEL DRAINS AND SOURCES AND
N ⁺ GUARD RINGS (top ohmic contact to substrate) | (MASK #5) |
| 33. | Develop photoresist | |
| 34. | n ⁺ IMPLANT | |
| 35. | Strip photoresist | |
| 36. | Strip thin oxide | |
| 37. | Grow oxide | |
| 38. | Apply photoresist | |
| 39. | PATTERN CONTACT OPENINGS | (MASK #6) |
| 40. | Develop photoresist | |
| 41. | Etch oxide | |
| 42. | Strip photoresist | |
| 43. | APPLY METAL | |
| 44. | Apply photoresist | |
| 45. | PATTERN METAL | (MASK #7) |
| 46. | Develop photoresist | |
| 47. | Etch metal | |

Typical Process Description (cont)

- 48. Strip photoresist
 - Optional steps for double metal process*
 - C.1 Strip thin oxide
 - C.2 DEPOSIT INTERMETAL OXIDE
 - C.3 Apply photoresist
 - C.4 PATTERN VIAS (MASK #C1)
 - C.5 Develop photoresist
 - C.6 Etch oxide
 - C.7 Strip photoresist
 - C.8 APPLY METAL (Metal 2)
 - C.9 Apply photoresist
 - C.10 PATTERN METAL (MASK #C2)
 - C.11 Develop photoresist
 - C.12 Etch metal
 - C.13 Strip photoresist
- 49. APPLY PASSIVATION
- 50. Apply photoresist
- 51. PATTERN PAD OPENINGS (MASK #8)
- 52. Develop photoresist
- 53. Etch passivation
- 54. Strip photoresist
- 55. ASSEMBLE, PACKAGE AND TEST

Typical Model Parameters

Process parameters for a typical^a p-well CMOS process

	Typical	Tolerance ^b	Units
Square law model parameters			
V_{T0} (threshold voltage)			
n-channel (V_{TN0})	0.75	± 0.25	V
p-channel (V_{TP0})	-0.75	± 0.25	V
K' (conduction factor)			
n-channel	24	± 6	$\mu\text{A}/\text{V}^2$
p-channel	8	± 1.5	$\mu\text{A}/\text{V}^2$
γ (body effect)			
n-channel	0.8	± 0.4	$\text{V}^{1/2}$
p-channel	0.4	± 0.2	$\text{V}^{1/2}$
λ (channel length modulation)			
n-channel	0.01	$\pm 50\%$	V^{-1}
p-channel	0.02	$\pm 50\%$	V^{-1}
ϕ (surface potential)			
n- and p-channel	0.6	± 0.1	V
Process parameters			
μ (channel mobility)			
n-channel	710		$\text{cm}^2/(\text{V} \cdot \text{s})$
p-channel	230		$\text{cm}^2/(\text{V} \cdot \text{s})$
Doping^c			
n^+ active	5	± 4	$10^{18}/\text{cm}^3$
p^+ active	5	± 4	$10^{17}/\text{cm}^3$
p-well	5	± 2	$10^{16}/\text{cm}^3$
n-substrate	1	± 0.1	$10^{16}/\text{cm}^3$

Typical Model Parameters (cont)

Physical feature sizes			
T_{ox} (gate oxide thickness)	500	± 100	\AA
Total lateral diffusion			
n-channel	0.45	± 0.15	μ
p-channel	0.6	± 0.3	μ
Diffusion depth			
n^+ diffusion	0.45	± 0.15	μ
p^+ diffusion	0.6	± 0.3	μ
p-well	3.0	$\pm 30\%$	μ
Insulating layer separation			
POLY I to POLY II	800	± 100	\AA
Metal 1 to Substrate	1.55	± 0.15	μ
Metal 1 to Diffusion	0.925	± 0.25	μ
POLY I to Substrate (POLY I on field oxide)	0.75	± 0.1	μ
Metal 1 to POLY I	0.87	± 0.7	μ
Metal 2 to Substrate	2.7	± 0.25	μ
Metal 2 to Metal I	1.2	± 0.1	μ
Metal 2 to POLY I	2.0	± 0.07	μ

Typical Model Parameters (cont)

Capacitances ^d			
C_{OX} (gate oxide capacitance, n- and p-channel)	0.7	± 0.1	fF/ μ^2
POLY I to substrate, poly in field	0.045	± 0.01	fF/ μ^2
POLY II to substrate, poly in field	0.045	± 0.01	fF/ μ^2
Metal 1 to substrate, metal in field	0.025	± 0.005	fF/ μ^2
Metal 2 to substrate, metal in field	0.014	± 0.002	fF/ μ^2
POLY I to POLY II	0.44	± 0.05	fF/ μ^2
POLY I to Metal 1	0.04	± 0.01	fF/ μ^2
POLY I to Metal 2	0.039	± 0.003	fF/ μ^2
Metal 1 to Metal 2	0.035	± 0.01	fF/ μ^2
Metal 1 to diffusion	0.04	± 0.01	fF/ μ^2
Metal 2 to diffusion	0.02	± 0.005	fF/ μ^2
n ⁺ diffusion to p-well (junction, bottom)	0.33	± 0.17	fF/ μ^2
n ⁺ diffusion sidewall (junction, sidewall)	2.6	± 0.6	fF/ μ
p ⁺ diffusion to substrate (junction, bottom)	0.38	± 0.12	fF/ μ^2
p ⁺ diffusion sidewall (junction, sidewall)	3.5	± 2.0	fF/ μ
p-well to substrate (junction, bottom)	0.2	± 0.1	fF/ μ^2
p-well sidewall (junction, sidewall)	1.6	± 1.0	fF/ μ
Resistances			
Substrate	25	$\pm 20\%$	$\Omega\text{-cm}$
p-well	5000	± 2500	Ω/\square
n ⁺ diffusion	35	± 25	Ω/\square
p ⁺ diffusion	80	± 55	Ω/\square
Metal	0.003	$\pm 25\%$	Ω/\square
Poly	25	$\pm 25\%$	Ω/\square
Metal 1–Metal 2 via ($3\ \mu \times 3\ \mu$ contact)	<0.1		Ω
Metal 1 contact to POLY I ($3\ \mu \times 3\ \mu$ contact)	<10		Ω
Metal 1 contact to n ⁺ or p ⁺ diffusion ($3\ \mu \times 3\ \mu$ contact)	<5		Ω

Typical Model Parameters (cont)

Breakdown voltages, leakage currents, migration currents and operating conditions

Punchthrough voltages (Gate oxide, POLY I to POLY II)	>10	V
Diffusion reverse breakdown voltage	>10	V
p-well to substrate reverse breakdown voltage	>20	V
Metal 1 in field threshold voltage	>10	V
Metal 2 in field threshold voltage	>10	V
Poly-field threshold voltage	>10	V
Maximum operating voltage	7.0	V
n ⁺ diffusion to p-well leakage current	0.25	fA/ μ^2
p ⁺ diffusion to substrate leakage current	0.25	fA/ μ^2
p-well leakage current	0.25	fA/ μ^2
Maximum metal current density	0.8	mA/ μ width
Maximum device operating temperature	200	°C

Typical Model Parameters (cont)

Level 3 Model (n-ch and p-ch)

SPICE MOSFET model parameters of a typical p-well CMOS process (MOSIS^a)

Parameter (Level 2 model)	n-channel	p-channel	Units
VTO	0.827	-0.895	V
KP	32.87	15.26	$\mu\text{A}/\text{V}^2$
GAMMA	1.36	0.879	$\text{V}^{1/2}$
PHI	0.6	0.6	V
LAMBDA	1.605E-2	4.709E-2	V^{-1}
CGSO	5.2E-4	4.0E-4	fF/ μ width
CGDO	5.2E-4	4.0E-4	fF/ μ width
RSH	25	95	Ω/\square
CJ	3.2E-4	2.0E-4	ρ fF/ μ^2
MJ	0.5	0.5	
CJSW	9.0E-4	4.5E-4	ρ fF/ μ perimeter
MJSW	0.33	0.33	
TOX	500	500	Å
NSUB	1.0E16	1.12E14	$1/\text{cm}^3$
NSS	0	0	$1/\text{cm}^2$
NFS	1.235E12	8.79E11	$1/\text{cm}^2$
TPG	1	-1	
XJ	0.4	0.4	μ
LD	0.28	0.28	μ
UO	200	100	$\text{cm}^2/(\text{V} \cdot \text{s})$
UCRIT	9.99E5	1.64E4	V/cm
UEXP	1.001E-3	0.1534	
VMAX	1.0E5	1.0E5	m/s
NEFF	1.001E-2	1.001E-2	
DELTA	1.2405	1.938	

Typical Model Parameters (cont) BSIM 4 Model (n-ch)

.MODEL CMOSN NMOS (LEVEL	= 49
+VERSION	= 3.1	TNOM	= 27
+XJ	= 1.5E-7	NCH	= 1.7E17
+K1	= 0.875093	K2	= -0.0943223
+K3B	= -8.5140476	WO	= 1.01582E-8
+DVTOW	= 0	DVT1W	= 0
+DVTO	= 2.670658	DVT1	= 0.4282172
+UO	= 452.3081836	UA	= 3.061716E-13
+UC	= 1.166279E-11	VSAT	= 1.682414E5
+AGS	= 0.1384489	BO	= 2.579158E-6
+KETA	= -3.615287E-3	A1	= 1.054571E-6
+RDSW	= 1.380341E3	PRWG	= 0.0301426
+WR	= 1	WINT	= 2.594349E-7
+XL	= 1E-7	XW	= 0
+DWB	= 3.537786E-8	VOFF	= 0
+CIT	= 0	CDSC	= 2.4E-4
+CDSCB	= 0	ETAO	= 2.332015E-3
+DSUB	= 0.076309	PCLM	= 2.6209353
+PDIBLC2	= 2.23243E-3	PDIBLCB	= -0.0436947
+PSCBE1	= 6.619472E8	PSCBE2	= 2.968801E-4
+DELTA	= 0.01	RSH	= 80.9
+PRT	= 0	UTE	= -1.5
+KT1L	= 0	KT2	= 0.022
+UB1	= -7.61E-18	UC1	= -5.6E-11
+WL	= 0	WLN	= 1
+WWN	= 1	WWL	= 0
+LLN	= 1	LW	= 0
+LWL	= 0	CAPMOD	= 2
+CGDO	= 2.34E-10	CGSO	= 2.34E-10
+CJ	= 4.240724E-4	PB	= 0.9148626
+CJSW	= 3.007134E-10	PBSW	= 0.8
+CJSWG	= 1.64E-10	PBSWG	= 0.8
+CF	= 0	PVTHO	= 0.0526696
+PK2	= -0.0283027	WKETA	= -0.0191754
		TOX	= 1.4E-8
		VTHO	= 0.6656437
		K3	= 25.0562441
		NLX	= 1E-9
		DVT2W	= 0
		DVT2	= -0.1373089
		UB	= 1.515137E-18
		AO	= 0.6297744
		B1	= 5E-6
		A2	= 0.3379035
		PRWB	= 0.0106493
		LINT	= 7.489566E-8
		DWG	= -9.471353E-9
		NFACTOR	= 1.0754804
		CDSCD	= 0
		ETAB	= -1.531255E-4
		PDIBLC1	= 1
		DROUT	= 1.0300278
		PVAG	= 9.970995E-3
		MOBMOD	= 1
		KT1	= -0.11
		UA1	= 4.31E-9
		AT	= 3.3E4
		WW	= 0
		LL	= 0
		LWN	= 1
		XPART	= 0.5
		CGBO	= 1E-9
		MJ	= 0.4416777
		MJSW	= 0.2025106
		MJSWG	= 0.2025106
		PRDSW	= 110.1539295
		LKETA	= 8.469064E-4

98 parameters in this BSIM Model !

Typical Model Parameters (cont)

BSIM 4 Model (p-ch)

```
.MODEL CMOSF PMOS (
+VERSION = 3.1          TNOM    = 27          LEVEL    = 49
+XJ      = 1.5E-7        NCH     = 1.7E17       TOX       = 1.4E-8
+K1       = 0.5600277    K2      = 9.302429E-3    VTH0      = -0.9633249
+K3B      = -1.0103515   WO       = 1.010628E-8    K3        = 7.2192028
+DVTOW    = 0           DVT1W    = 0           NLX       = 5.826683E-8
+DVTO     = 2.2199372   DVT1     = 0.5378964    DVT2W     = 0
+UO       = 220.5729225  UA       = 3.141811E-9    DVT2      = -0.1158128
+UC       = -5.76898E-11 VSAT    = 1.342779E5    UB        = 1.085892E-21
+AGS      = 0.157364    BO       = 9.735259E-7    AO        = 0.9333822
+KETA     = -2.42686E-3  A1      = 3.447019E-4    B1        = 5E-6
+RDSW     = 3E3         PRWG     = -0.0418484    A2        = 0.3701317
+WR       = 1           WINT     = 3.097872E-7    PRWB      = -0.0212357
+XL       = 1E-7        XW      = 0           LINT      = 1.040878E-7
+DWB      = 1.629532E-8 VOFF    = -0.0823738    DWG       = -1.983686E-8
+CIT       = 0          CDSC     = 2.4E-4       NFACTOR   = 0.969384
+CDSCB    = 0          ETAO     = 0.4985496    CDSCD     = 0
+DSUB     = 1          PCLM     = 2.1142057    ETAB      = -0.0653358
+PDIBLC2  = 3.172604E-3 PDIBLCB = -0.0511673    PDIBLC1   = 0.0256688
+PSCBE1   = 1.851867E10 PSCBE2  = 1.697939E-9    DROUT     = 0.1695622
+DELTA    = 0.01       RSH     = 103.6       PVAG      = 0
+PRT       = 0         UTE      = -1.5       MOBMOD    = 1
+KT1L     = 0         KT2      = 0.022    KT1       = -0.11
+UB1      = -7.61E-18 UC1      = -5.6E-11    UA1       = 4.31E-9
+WL       = 0         WLN      = 1         AT        = 3.3E4
+WWN      = 1         WWL      = 0         WW        = 0
+LLN      = 1         LW       = 0         LL        = 0
+LWL      = 0         CAPMOD   = 2         LWN       = 1
+CGDO     = 3.09E-10   CGSO    = 3.09E-10    XPART     = 0.5
+CJ       = 7.410008E-4 PB       = 0.9665307   CGBO      = 1E-9
+CJSW     = 2.487127E-10 PBSW    = 0.99        MJ        = 0.4978642
+CJSWG    = 6.4E-11   PBSWG   = 0.99        MJSW      = 0.3877813
+CF       = 0         PVTHO    = 5.98016E-3    MJSWG     = 0.3877813
+PK2      = 3.73981E-3 WKETA    = 2.870507E-3    PRDSW     = 14.8598424
-          -          -          -          LKETA     = -4.823171E-3
```



Stay Safe and Stay Healthy !

End of Lecture 7